

# **A VCO-Based ADC Employing a Multi-Phase Noise-Shaping Beat Frequency Quantizer for Direct Sampling of Sub-1mV Input Signals**

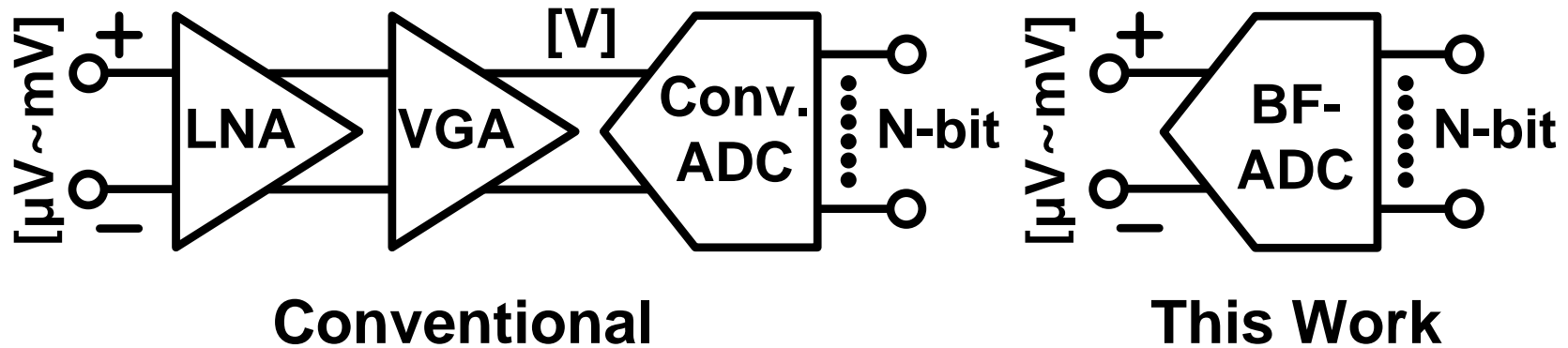
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# Agenda

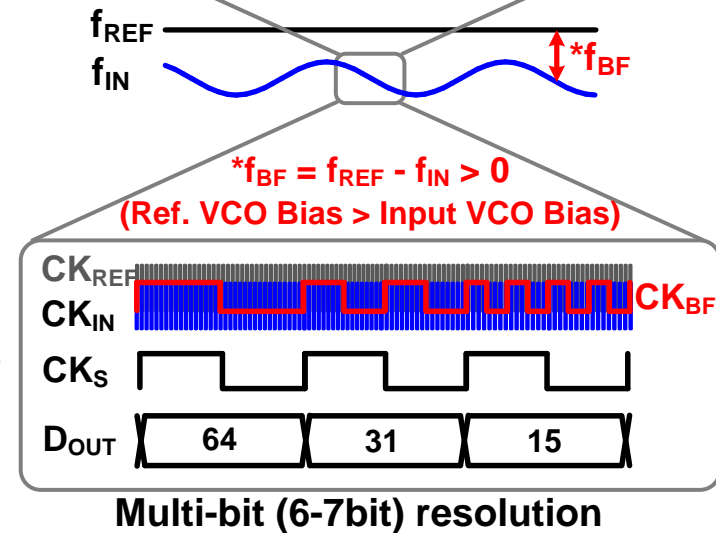
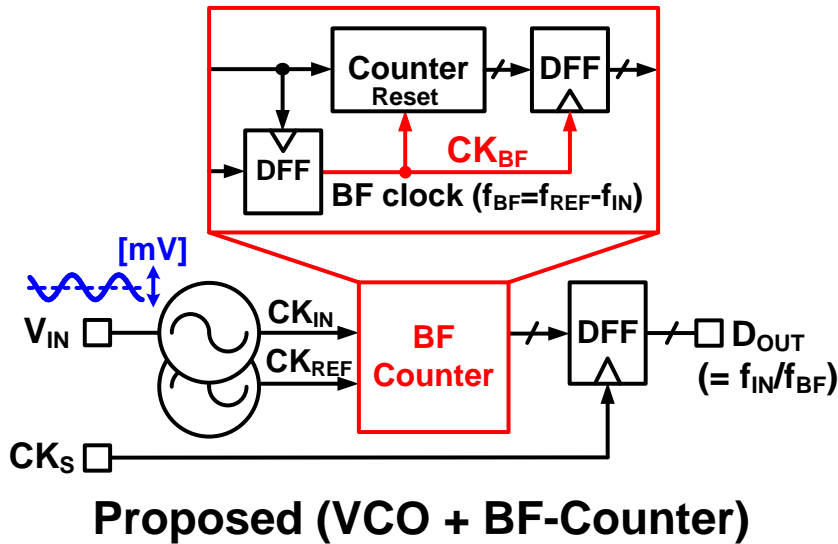
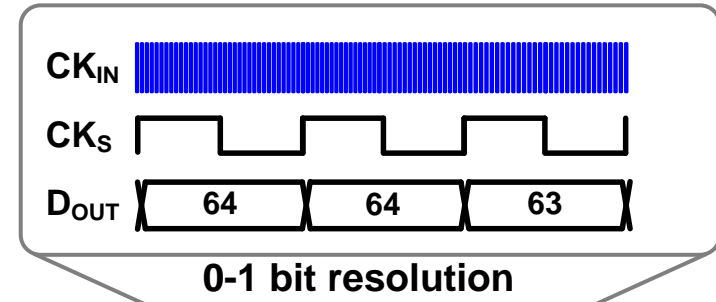
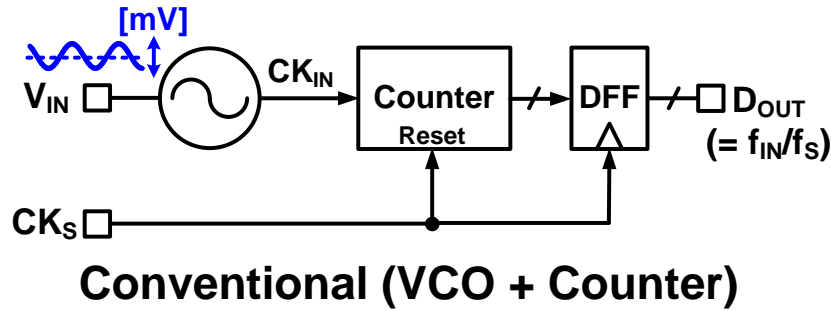
- **Motivation**
- **Conventional vs. proposed VCO-based ADC**
- **Multi-phase & Noise-shaping BF Quantizer**
- **Test chip configuration**
- **Measurement results**
- **Conclusion**

# Direct Sub-mV Signal Acquisition



- **Conventional**
  - Signal pre-conditioning amplifiers (LNA/VGA)
  - Conventional ADC for rail-to-rail input range
- **This work (Beat Frequency (BF) ADC)**
  - Direct A-to-D conversion w/o signal amplification

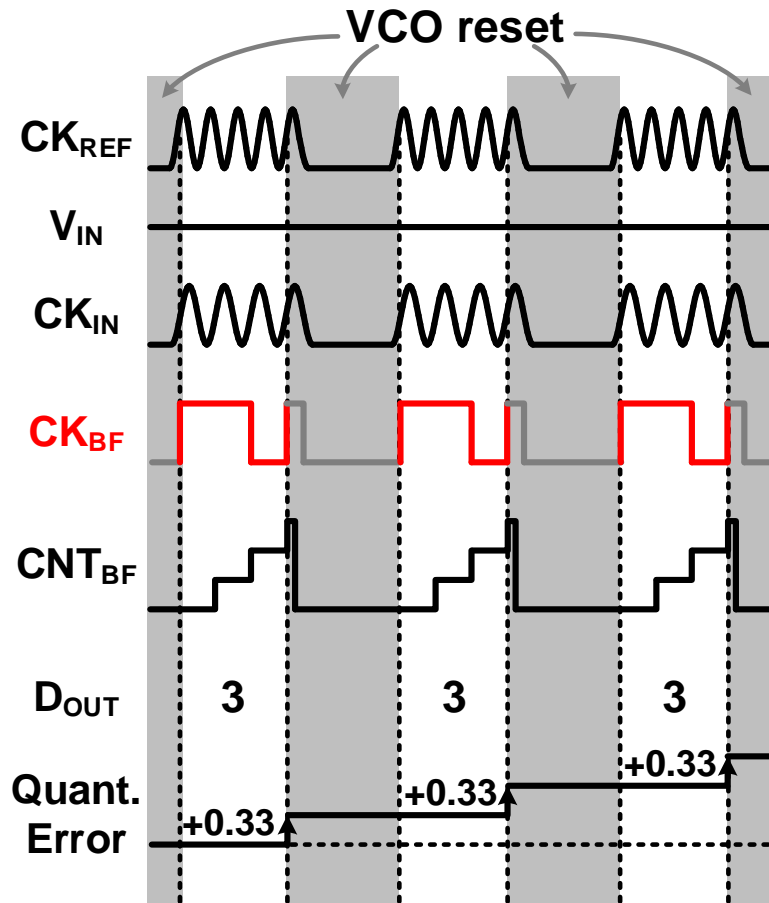
# Conv. vs. Proposed VCO-Based ADC



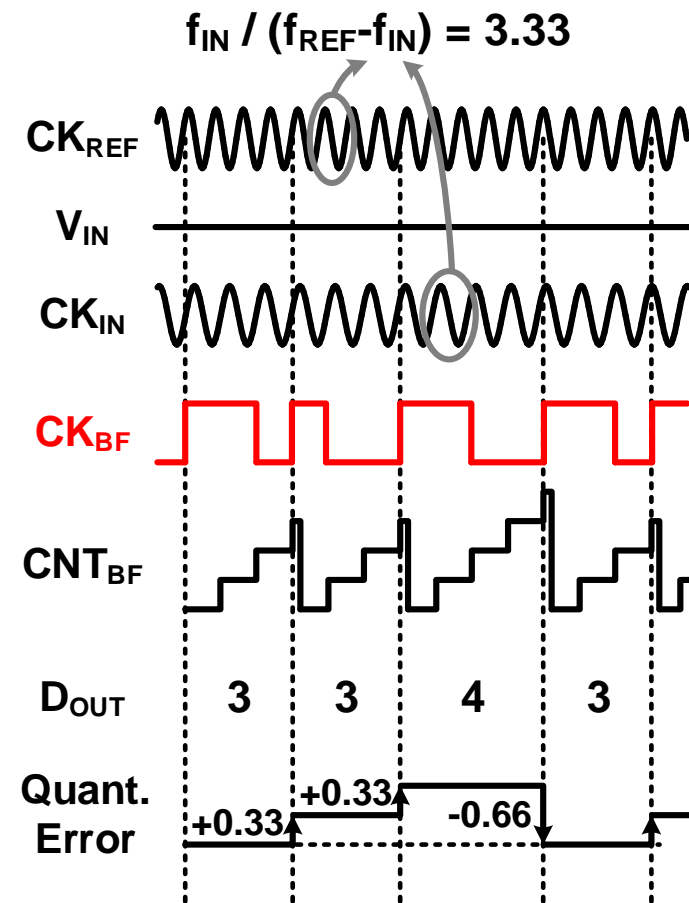
- Conventional : Linear freq. detection using VCO
- Proposed : High-resolution BF detection based [1]

[1] B. Kim et al., IEEE Custom Integrated Circuits Conference 2013

# Noise-Shaping Beat Freq. Quant.



Prior work [1]: Nyquist rate

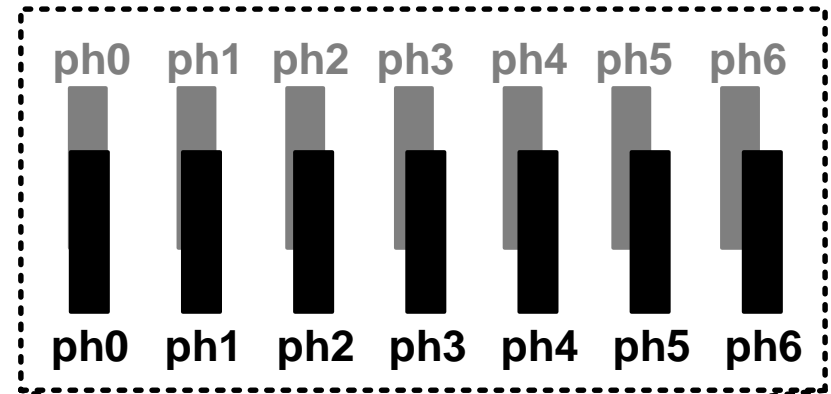
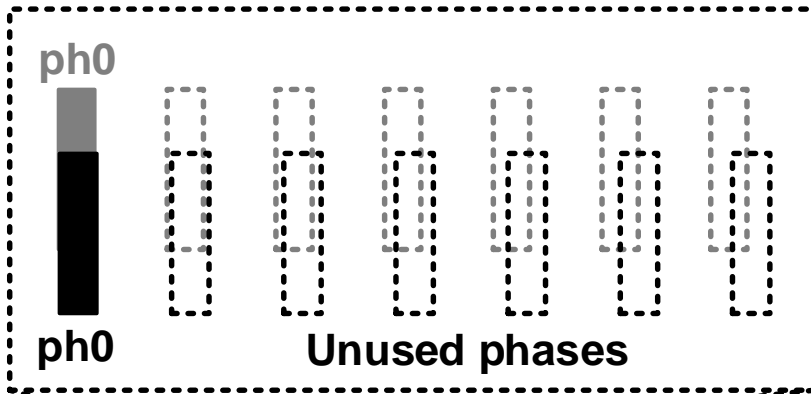


This work: Noise-shaping

- Noise-shaping improves BF detection resolution

# Multi-Phase Beat Freq. Quantization

(Example of a 7-phase ring oscillator)



(Single-phase resolution)

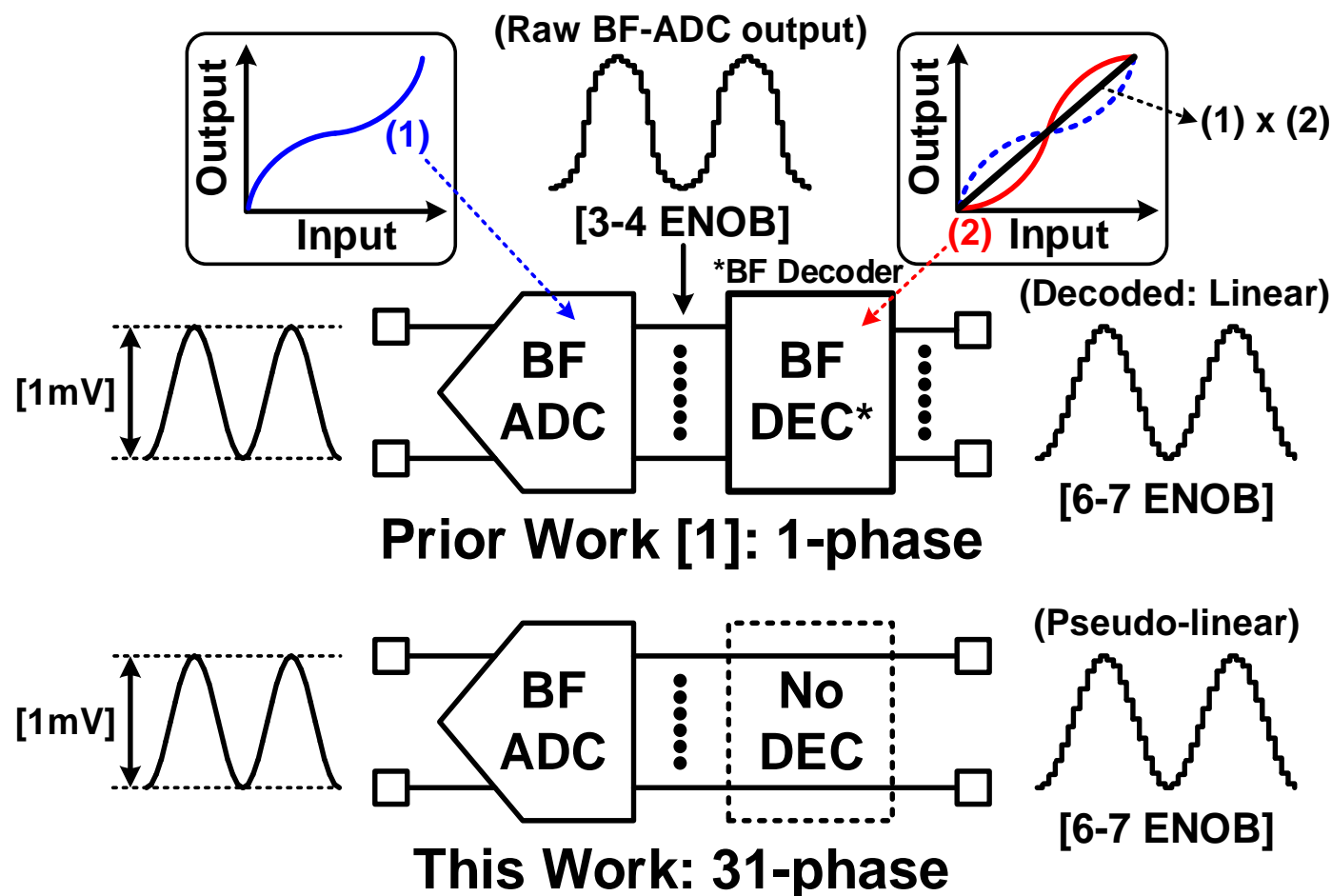
Prior work [1]

(Multi-phase resolution)

This work

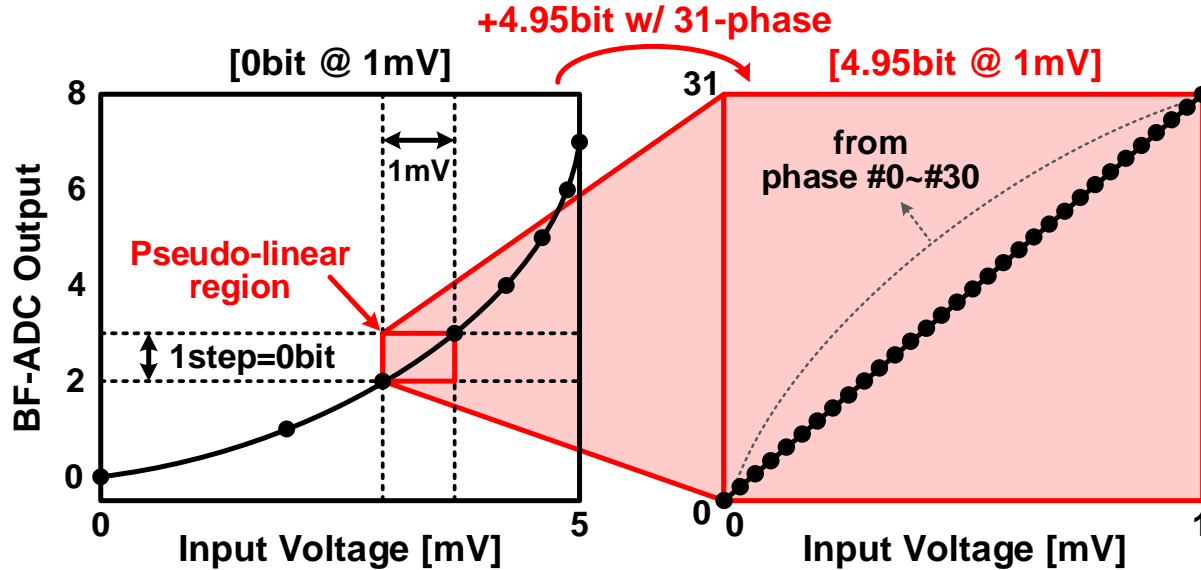
- N-phase improves BF detection resolution N-times

# Linear BF-ADC with Multi-Phase



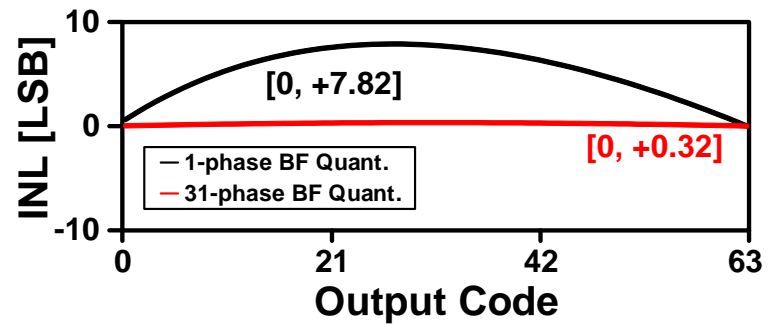
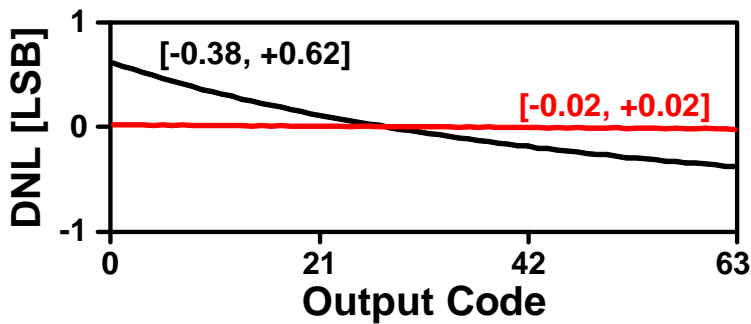
- Does not require BF-decoding with the improved resolution from multi-phase

# Linearity of Multi-Phase BF-ADC



1-Phase BF Quantizer (#0)

31-Phase BF Quantizer



- Simulated linearity:  $DNL_{MAX}/INL_{MAX}=0.02/0.32$  [LSB]

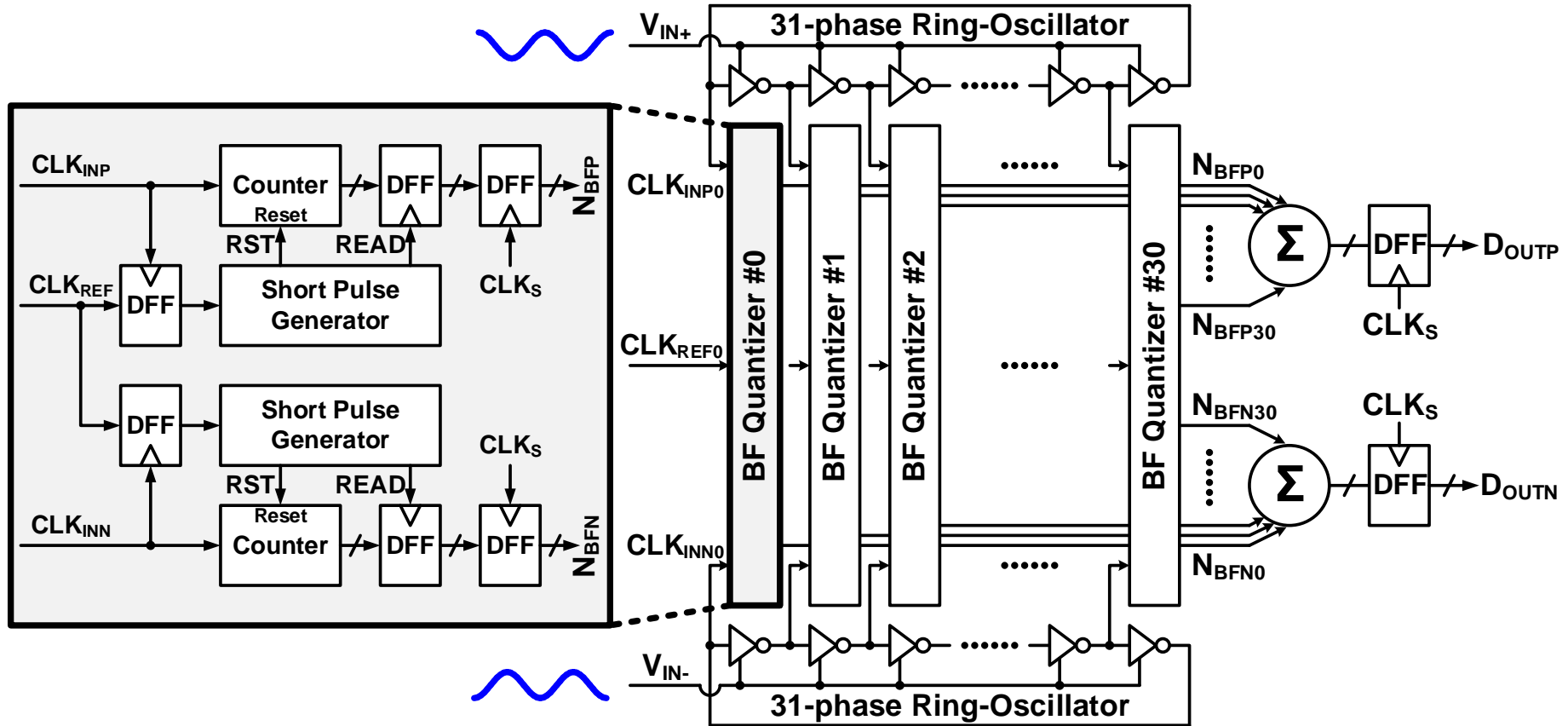


# Comparison with Prior Arts

	Conventional VCO-based	Previous BF-ADC [1]	This work
Quantization Scheme	Linear Counting	Beat Freq. Counting	Beat Freq. Counting
Freq. Detection Sensitivity* (@ 100 counts)	Low (1%)	High (0.01%)	High (0.01%)
Noise-Shaping	Yes	No	Yes
Multi-Phase	Yes	No	Yes
BF Code Utilization	-	Wide (Nonlinear)	Narrow (Pseudo-linear)
BF Decoding	-	Required	Not Required

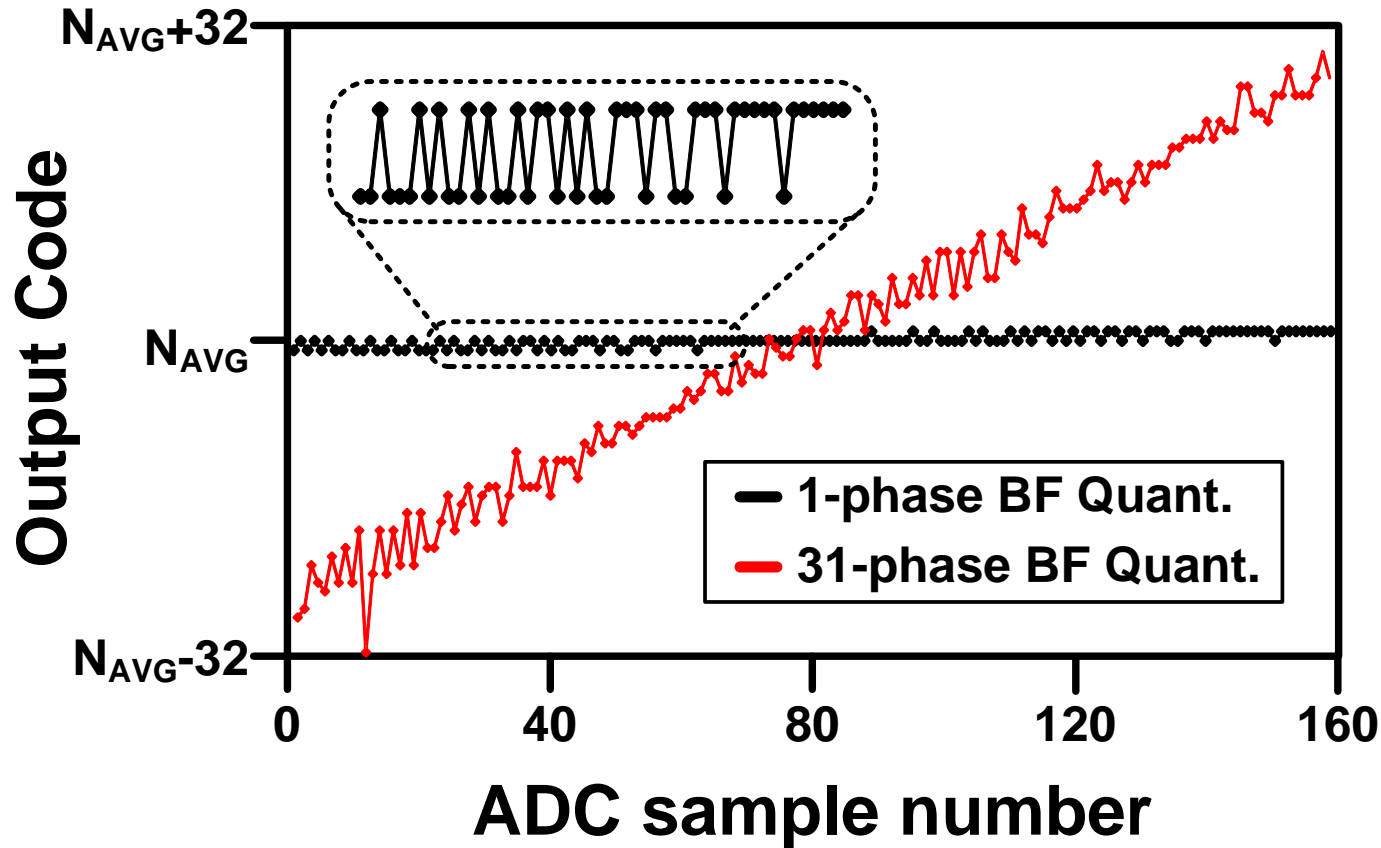
\*Freq. step required for a count change of one

# VCO-Based ADC w/ BF-Quantizer



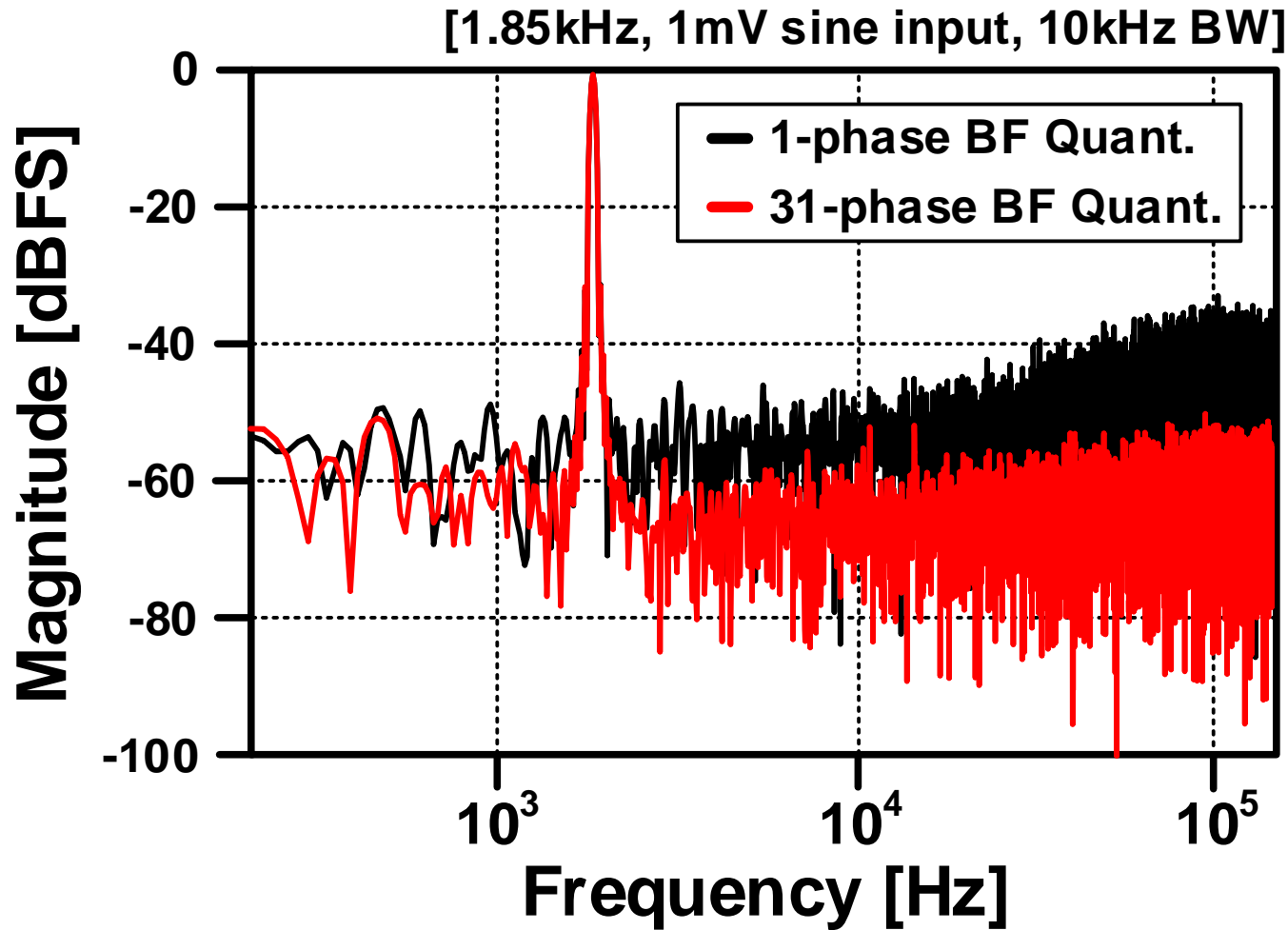
- A pair of 31-phase ring oscillator for +/- inputs
- 31x banks (each consists of +/- BF quantizers)

# Measured Code vs. Ramp Input



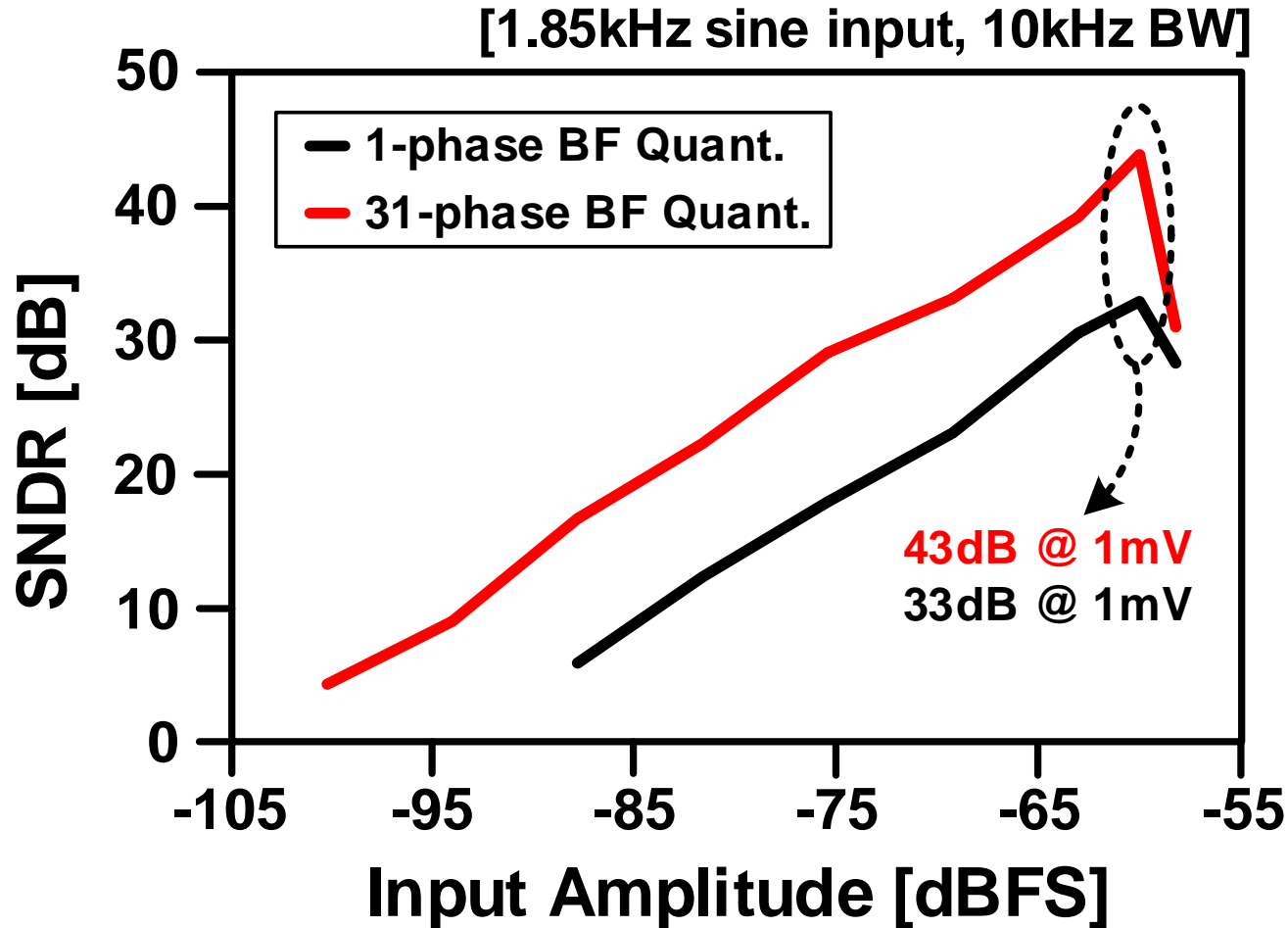
- Noise-shaping is clearly shown in the results
- 31-phase result shows 31x wider output range

# Measured FFT Result w/ 1mV Input



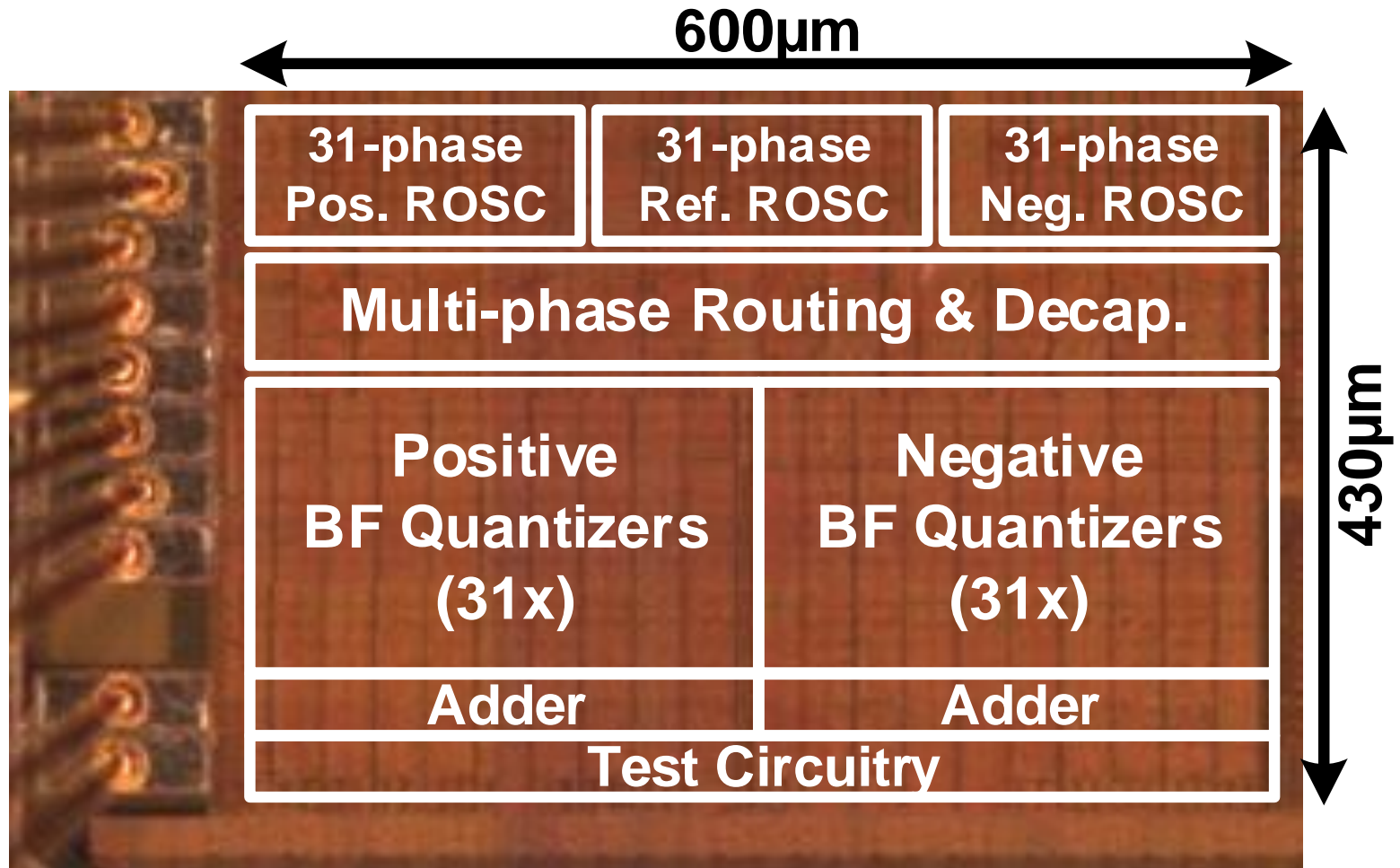
- SNDR = 43dB (i.e. ENOB=6.85) / SFDR = 56.7dB

# Measured SNDR vs. Input Amplitude



- SNDR @ 1mV = 43dB @ 31-phase, 33dB @ 1-phase

# Chip Micrograph



# Performance Comparison

	VLSI'07	VLSI'11	[1] CICC'13	This work
<b>Process</b>	0.13 $\mu$ m	90nm	65nm	65nm
<b>Supply</b>	1.2V	N/A	1.2V	1.2V
<b>Sample Rate</b>	950MHz	640MHz	4.17kHz	300kHz
<b>Input BW</b>	20MHz	8MHz	2kHz	10kHz
<b>SNDR<sub>1mV</sub>*</b>	12dB	3dB	35dB	43dB
<b>ENOB<sub>1mV</sub>*</b>	1.70	0.21	5.52	6.85
<b>SFDR[dB]</b>	N/A	71.4	41.9	56.7
<b>IN<sub>0dB</sub>[dBFS]**</b>	-70	-63	-89	-105
<b>Power</b>	38mW	4.3mW	0.92 $\mu$ W	36 $\mu$ W
<b>Area[mm<sup>2</sup>]</b>	0.185	0.10	0.013	0.258

\* Peak SNDR/ENOB for a 1mV input amplitude.

\*\* Input amplitude at SNDR = 0dB (dBFS @ full-scale = 1V)

# Conclusion

- **A VCO-based ADC featuring beat-freq. quantizer is designed using 65nm CMOS for a direct A-to-D conversion of sub-mV input signal**
- **Multi-phase and 1<sup>st</sup> order noise-shaping property improves ADC resolution without the use of beat frequency decoding**
- **43dB SNDR (6.85 ENOB) has been achieved for a 1mVppd input signal**