

## Linear State-Space Equations

$$\begin{bmatrix} \dot{x}_1(t) \\ \vdots \\ \dot{x}_n(t) \end{bmatrix} = \underbrace{\begin{bmatrix} a_{11} & \dots & a_{1n} \\ \vdots & & \vdots \\ a_{n1} & \dots & a_{nn} \end{bmatrix}}_A \begin{bmatrix} x_1(t) \\ \vdots \\ x_n(t) \end{bmatrix} + \underbrace{\begin{bmatrix} b_{11} & \dots & b_{1p} \\ \vdots & & \vdots \\ b_{n1} & \dots & b_{np} \end{bmatrix}}_B \begin{bmatrix} u_1(t) \\ \vdots \\ u_p(t) \end{bmatrix}$$

$x_1(t), \dots, x_n(t)$  are state-variables  
 $u_1(t), \dots, u_p(t)$  are system inputs

We can define  $\underline{x}(t) := \begin{bmatrix} x_1(t) \\ \vdots \\ x_n(t) \end{bmatrix}$        $\underline{u}(t) := \begin{bmatrix} u_1(t) \\ \vdots \\ u_p(t) \end{bmatrix}$

$$\dot{\underline{x}}(t) = A \underline{x}(t) + B \underline{u}(t) \quad (\text{State-Equations})$$

State equations are n linear coupled 1st order diff. eq'ns

The outputs are linear combinations of the state-variables and the inputs.

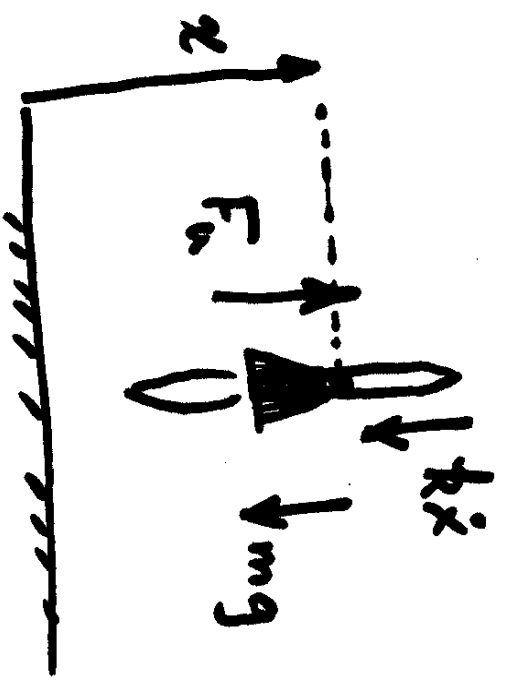
$$\underbrace{\begin{bmatrix} y_1(t) \\ \vdots \\ y_q(t) \end{bmatrix}}_{\underline{y}(t)} = \underbrace{\begin{bmatrix} c_{11} & \dots & c_{1n} \\ \vdots & & \vdots \\ c_{q1} & \dots & c_{qn} \end{bmatrix}}_C \underbrace{\begin{bmatrix} x_1(t) \\ \vdots \\ x_n(t) \end{bmatrix}}_{\underline{x}} + \underbrace{\begin{bmatrix} d_{11} & \dots & d_{1p} \\ \vdots & & \vdots \\ d_{q1} & \dots & d_{qp} \end{bmatrix}}_D \underbrace{\begin{bmatrix} u_1(t) \\ \vdots \\ u_p(t) \end{bmatrix}}_{\underline{u}}$$

$$\underline{y}(t) = C \underline{x}(t) + D \underline{u}(t) \quad (\text{Output-Equations})$$

Together, state equations & output equations are the state-space description of the system

$$\begin{aligned} \dot{\underline{x}}(t) &= A \underline{x}(t) + B \underline{u}(t) \\ \underline{y}(t) &= C \underline{x}(t) + D \underline{u}(t) \end{aligned}$$

# Example 1 (Rocket)



$F_u$  Thrust force  
 $mg$  Weight of rocket  
 $k\dot{x}$  Air friction force

Newton's law:  $F_u - mg - k\dot{x} = m\ddot{x}$

or  $m\ddot{x} + k\dot{x} = (F_u - mg)$

2nd order linear  
diff. eq'n

dividing by  $m$   $\ddot{x} + \frac{k}{m}\dot{x} = \frac{F_u}{m} - g$

Let  $x_1 := x$  (Rocket position)

$x_2 := \dot{x}$  (Rocket velocity)

$$\dot{x}_1(t) = x_2(t)$$

$$\dot{x}_2(t) = \ddot{x}(t)$$

$$= -\frac{k}{m} \ddot{x}(t) + \frac{F_0}{m} - g$$

$$= -\frac{k}{m} x_2(t) + \frac{F_0}{m} - g$$

In matrix form:

$$\underbrace{\begin{bmatrix} \dot{x}_1(t) \\ \dot{x}_2(t) \end{bmatrix}}_{\dot{\underline{x}}(t)} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & -\frac{k}{m} \end{bmatrix}}_A \underbrace{\begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}}_{\underline{x}} + \underbrace{\begin{bmatrix} 0 \\ 1 \end{bmatrix}}_B \underbrace{\left(\frac{F_0}{m} - g\right)}_{u(t)}$$

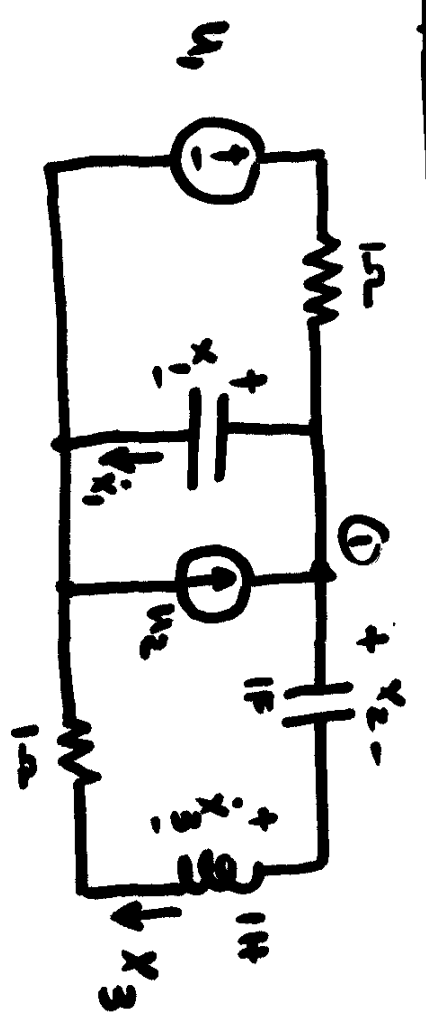
$$y(t) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}$$

if velocity is output of interest

$$\text{or } y(t) = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix}$$

if position is output of interest

# Example 2 (Electric Circuit)



Use: Inductor currents  
 Capacitor voltages  
 as state-variables

Summing current at node ①

$$(u_1 - x_1) - i_1 + u_2 - x_3 = 0 \quad \dots \quad (I)$$

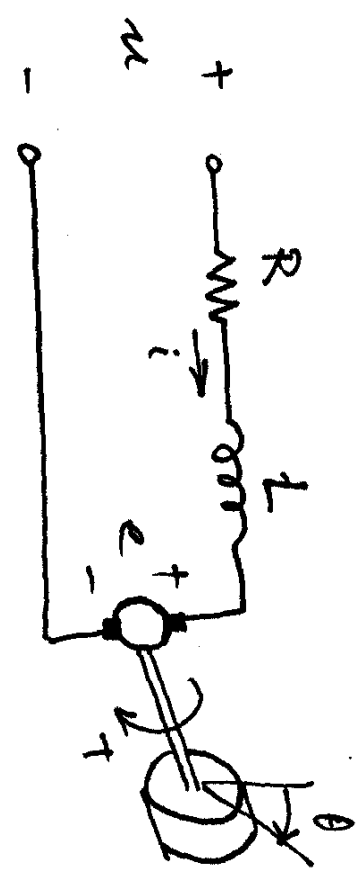
$$\dot{x}_2 = x_3 \quad \dots \quad (II)$$

$$\dot{x}_3 + x_3 - x_1 + x_2 = 0 \quad \dots \quad (III)$$

In matrix form:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} -1 & 0 & 0 \\ 0 & 1 & -1 \\ 1 & -1 & -1 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 1 & 1 \\ 0 & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} u_1 \\ u_2 \end{bmatrix}$$

# Example 3 (Electric Motor)



$$u = Ri + L \frac{di}{dt} + e \Rightarrow \frac{di}{dt} = -\frac{R}{L}i - \frac{e}{L} + \frac{u}{L}$$

$$T = I \ddot{\theta} \Rightarrow \ddot{\theta} = \frac{k_1}{I} i$$

$$e = k_2 \dot{\theta}$$

Let  $x_1 = \theta$   
 $x_2 = \dot{\theta}$   
 $x_3 = \ddot{\theta}$

Then

$$\begin{aligned} \dot{x}_1 &= x_2 \\ \dot{x}_2 &= \frac{k_1}{I} x_3 \\ \dot{x}_3 &= -\frac{R}{L} x_3 - \frac{1}{L} k_2 x_2 + \frac{u}{L} \end{aligned}$$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & \frac{k_1}{I} \\ 0 & -\frac{k_2}{L} & -\frac{R}{L} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \frac{1}{L} \end{bmatrix} u$$

If  $y = \theta$

$$y = \begin{bmatrix} 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + 0 \cdot u$$

### Example 4

$$\overset{\circ\circ\circ}{y}(t) + a_2 \overset{\circ\circ}{y}(t) + a_1 \overset{\circ}{y}(t) + a_0 y(t) = \tilde{u}(t)$$

Let  $x_1 := y$

$$x_2 = \overset{\circ}{y}$$

$$x_3 = \overset{\circ\circ}{y}$$

Then  $\overset{\circ}{x}_1 = x_2$

$$\overset{\circ}{x}_2 = x_3$$

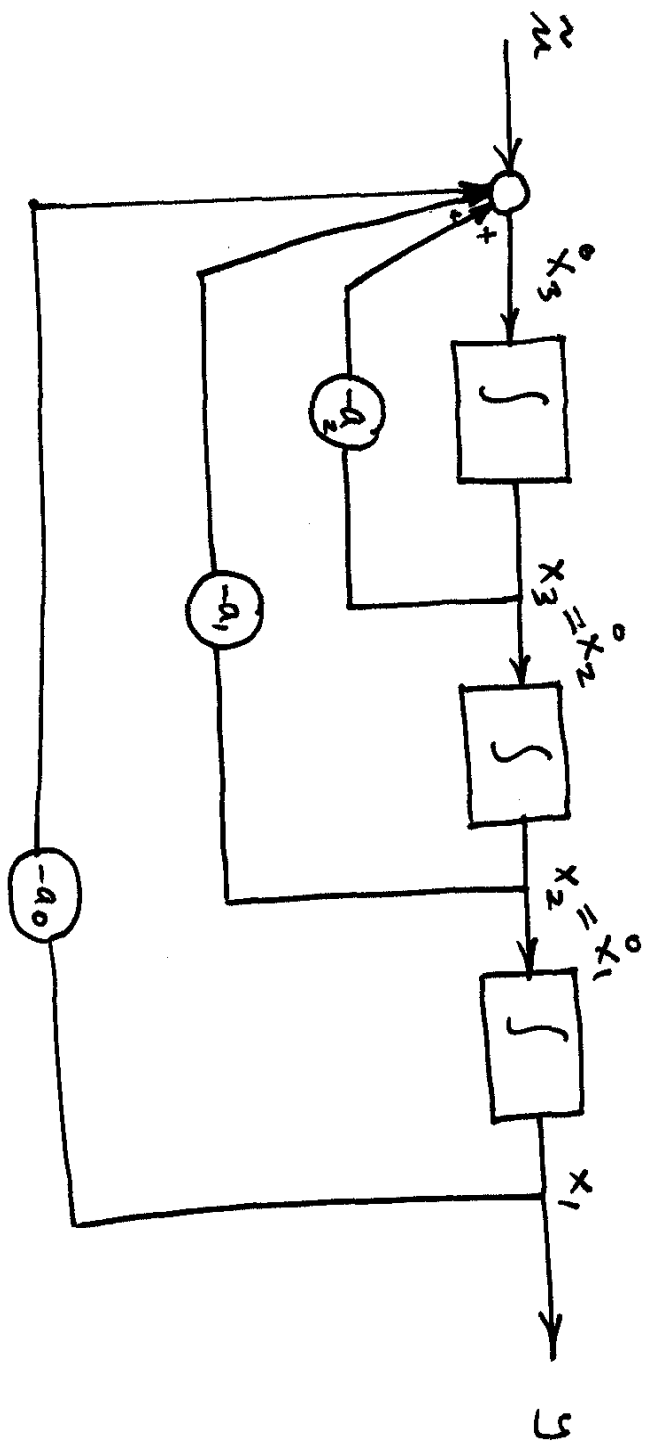
$$\overset{\circ\circ}{x}_3 = \overset{\circ\circ\circ}{y}(t) = -a_2 \overset{\circ\circ}{y}(t) - a_1 \overset{\circ}{y}(t) - a_0 y(t) + \tilde{u}(t)$$

$$= -a_2 x_3 - a_1 x_2 - a_0 x_1 + \tilde{u}$$

$$\begin{bmatrix} \overset{\circ}{x}_1 \\ \overset{\circ}{x}_2 \\ \overset{\circ\circ}{x}_3 \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 \end{bmatrix}}_A \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}}_B \tilde{u} + \underbrace{\begin{bmatrix} 0 \\ 0 \\ 0 \end{bmatrix}}_D \tilde{u}$$

$$y = \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$

# Analog Computer Simulation





Example 5 :

$$y(s) = \frac{b_2 s^2 + b_1 s + b_0}{s^3 + a_2 s^2 + a_1 s + a_0} u(s)$$

Or equivalently,  $\overset{\circ\circ}{y}(t) + a_2 \overset{\circ\circ}{y}(t) + a_1 \overset{\circ}{y}(t) + a_0 y(t) = b_2 \overset{\circ\circ}{u}(t) + b_1 \overset{\circ}{u}(t) + b_0 u(t)$

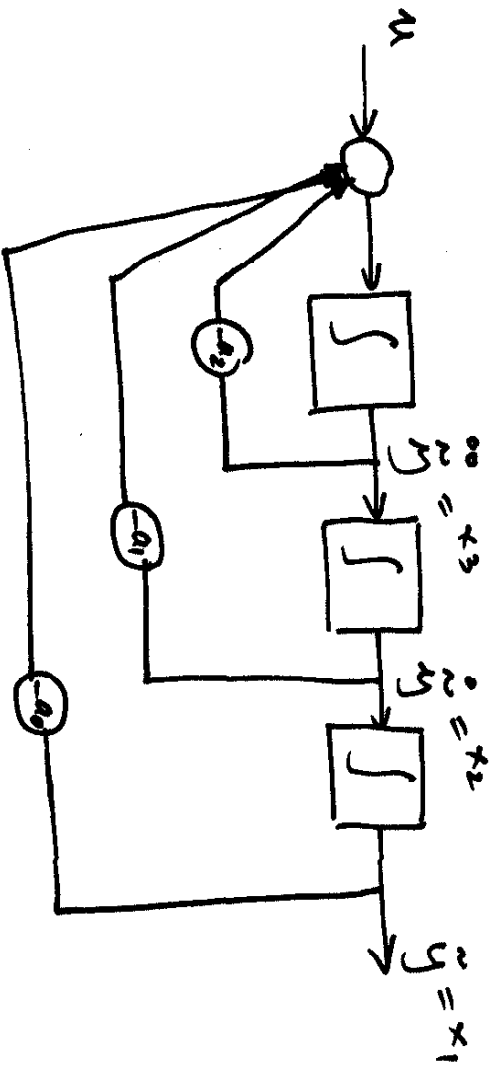
To get a state-space representation, we solve the problem in 2 steps:

① Letting  $\overset{\sim}{y}(s) ::= \frac{u(s)}{s^3 + a_2 s^2 + a_1 s + a_0}$ , we can obtain a state-space

description with  $\overset{\sim}{y}(s)$  as the output:

$$(s^3 + a_2 s^2 + a_1 s + a_0) \overset{\sim}{y}(s) = u(s) \quad \sigma$$

$$\overset{\circ\circ}{\ddot{y}}(t) + a_2 \overset{\circ\circ}{\ddot{y}}(t) + a_1 \overset{\circ}{\dot{y}}(t) + a_0 \overset{\sim}{y}(t) = u(t)$$



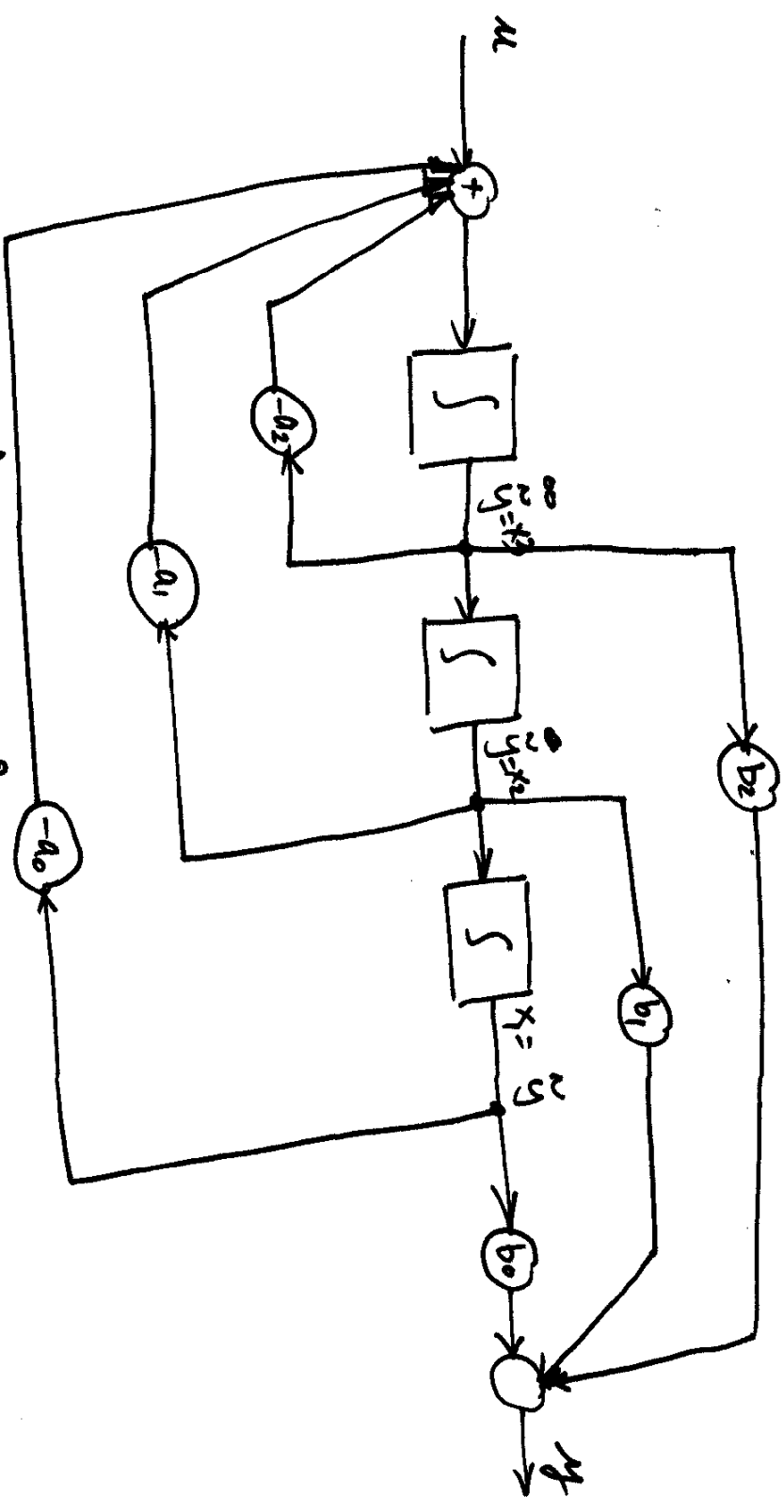
$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u$$

②  $y(s) = (b_2 s^2 + b_1 s + b_0) \tilde{y}(s) \sigma_z$

$$y(t) = b_2 \ddot{y}(t) + b_1 \dot{y}(t) + b_0 y(t)$$

$$= b_2 x_3 + b_1 x_2 + b_0 x_1$$

$$y(t) = \underbrace{[b_0 \quad b_1 \quad b_2]}_C \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix}$$



$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ -a_0 & -a_1 & -a_2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix} u$$

$$y = \underbrace{\begin{bmatrix} b_0 & b_1 & b_2 \end{bmatrix}}_D \begin{bmatrix} x_1 \\ x_2 \\ x_3 \end{bmatrix} + \underbrace{[0]}_D u$$

# Linear Systems as approximations to nonlinear Systems

Given:

$$\begin{bmatrix} \dot{x}_1 \\ \vdots \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} f_1(x_1, \dots, x_n, u_1, \dots, u_p) \\ \vdots \\ f_n(x_1, \dots, x_n, u_1, \dots, u_p) \end{bmatrix} \quad (\text{State eq's})$$

$$\begin{bmatrix} y_1 \\ \vdots \\ y_q \end{bmatrix} = \begin{bmatrix} g_1(x_1, \dots, x_n, u_1, \dots, u_p) \\ \vdots \\ g_q(x_1, \dots, x_n, u_1, \dots, u_p) \end{bmatrix} \quad (\text{Output eq's})$$

or

$$\dot{\underline{x}} = \underline{f}(\underline{x}, \underline{u}) \quad (\text{State eq's})$$

$$\underline{y} = \underline{g}(\underline{x}, \underline{u}) \quad (\text{Output eq's})$$

$(\underline{x}_0, \underline{u}_0) \in \mathbb{R}^n$  is an equilibrium solution if

$$f(\underline{x}_0, \underline{u}_0) = 0$$

In this case  $\underline{x}(t) \equiv \underline{x}_0$  and  $\underline{u}(t) \equiv \underline{u}_0$   
solve the differential equation (Constant Solution)

We can look at solutions close the equilibrium

$$\begin{aligned} \text{Let } \underline{x}(t) &::= \underline{x}_0 + \delta \underline{x}(t) \\ \underline{u}(t) &::= \underline{u}_0 + \delta \underline{u}(t) \end{aligned}$$

$$\dot{x}_i^0(t) = \delta \dot{x}_i = f_i(x, u)$$

$$= \cancel{f_i(x_0, u_0)} + \frac{\partial f_i}{\partial x_1} \Big|_{x_0, u_0} \delta x_1 + \dots + \frac{\partial f_i}{\partial x_n} \Big|_{x_0, u_0} \delta x_n$$

$$+ \frac{\partial f_i}{\partial u_1} \Big|_{x_0, u_0} \delta u_1 + \dots + \frac{\partial f_i}{\partial u_p} \Big|_{x_0, u_0} \delta u_p$$

+ higher order terms

Similarly for each output  $y_i \dots$

In matrix form:

$$\begin{bmatrix} \delta x_1 \\ \vdots \\ \delta x_n \end{bmatrix} = \underbrace{\begin{bmatrix} \frac{\partial f_1}{\partial x_1} & \dots & \frac{\partial f_1}{\partial x_n} \\ \vdots & & \vdots \\ \frac{\partial f_n}{\partial x_1} & \dots & \frac{\partial f_n}{\partial x_n} \end{bmatrix}}_A \begin{bmatrix} \delta x_1 \\ \vdots \\ \delta x_n \end{bmatrix} + \underbrace{\begin{bmatrix} \frac{\partial f_1}{\partial u_1} & \dots & \frac{\partial f_1}{\partial u_p} \\ \vdots & & \vdots \\ \frac{\partial f_n}{\partial u_1} & \dots & \frac{\partial f_n}{\partial u_p} \end{bmatrix}}_B \begin{bmatrix} \delta u_1 \\ \vdots \\ \delta u_p \end{bmatrix} + \text{higher order terms}$$

Ignoring H.O.T. we have

$$\underline{\delta x} \approx A \underline{\delta x} + B \underline{\delta u} \quad (\text{Linearized Input eqs})$$

$$\underline{\delta y} \approx C \underline{\delta x} + D \underline{\delta u} \quad (\text{Linearized Output eqs})$$