

# Sparsity-Promoting Optimal Control of Distributed Systems

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joint work with:

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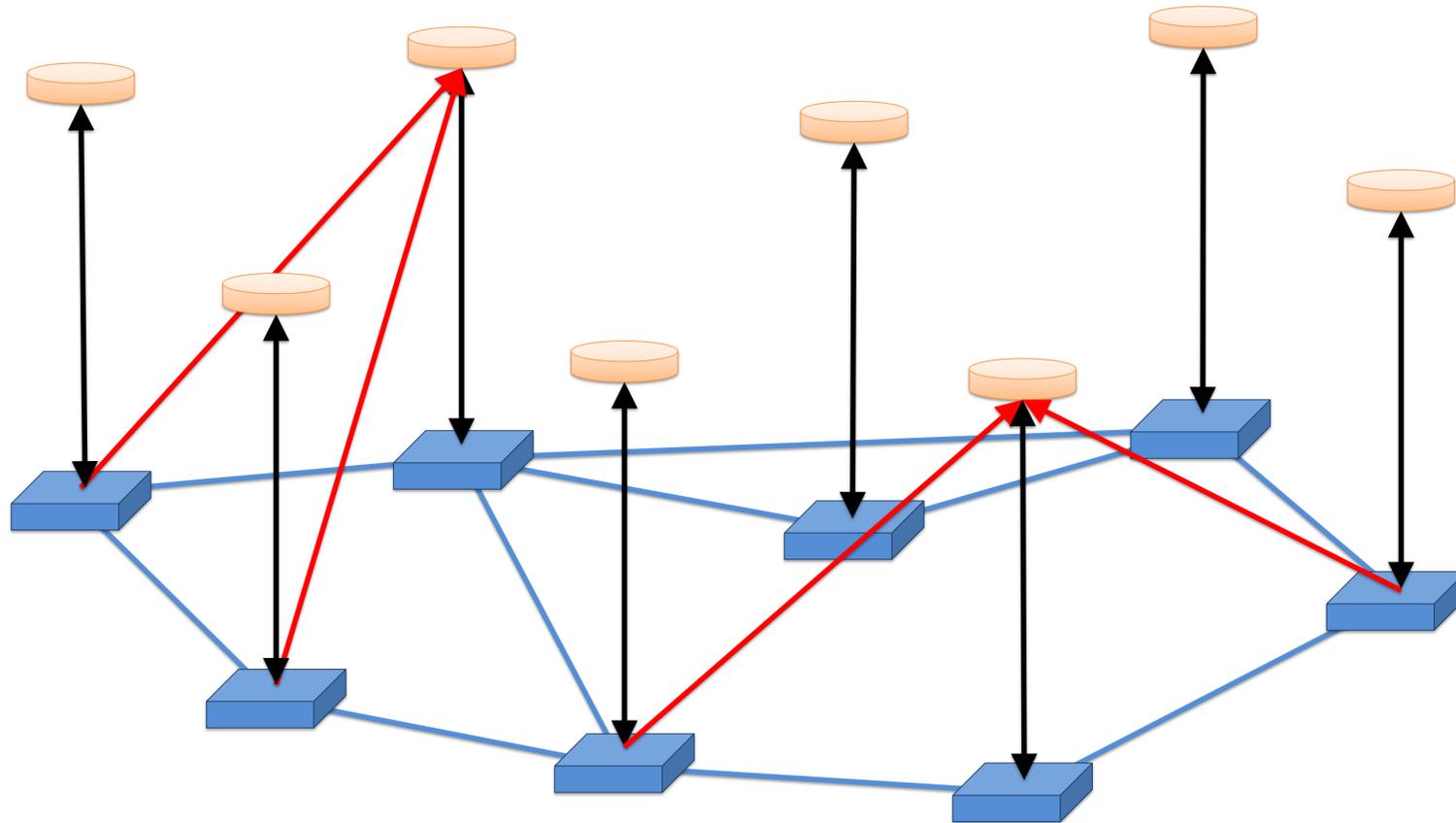
UNIVERSITY  
OF MINNESOTA

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# Structured distributed control

- **Blue layer:** distributed plant and its interaction links

structured memoryless controller



KEY CHALLENGE:

identification of a **signal exchange network**  
performance vs sparsity

# Minimum variance state-feedback problem

dynamics:  $\dot{x} = Ax + B_1 d + B_2 u$

objective function:  $J = \lim_{t \rightarrow \infty} \mathcal{E} (x^T(t) Q x(t) + u^T(t) R u(t))$

memoryless controller:  $u = -F x$

## • CLOSED-LOOP VARIANCE AMPLIFICATION

$$J(F) = \text{trace} \left( \int_0^{\infty} e^{(A - B_2 F)^T t} (Q + F^T R F) e^{(A - B_2 F) t} dt B_1 B_1^T \right)$$

★ no structural constraints

**globally optimal controller:**

$$\begin{aligned} A^T P + P A - P B_2 R^{-1} B_2^T P + Q &= 0 \\ F_c &= R^{-1} B_2^T P \end{aligned}$$

## SDP formulation

$$\underset{X, F}{\text{minimize}} \quad \text{trace} \left( (Q + F^T R F) X \right)$$

$$\text{subject to} \quad (A - B_2 F) X + X (A - B_2 F)^T = -B_1 B_1^T$$

$$X \succ 0$$

- CHANGE OF VARIABLES:  $FX = Y$

$$\underset{X, Y \succ 0}{\text{minimize}} \quad \text{trace}(Q X) + \text{trace}(R Y X^{-1} Y^T)$$

$$\text{subject to} \quad (A X - B_2 Y) + (A X - B_2 Y)^T + B_1 B_1^T = 0$$

## SDP characterization:

$$\underset{X, Y, Z}{\text{minimize}} \quad \text{trace}(Q X) + \text{trace}(R Z)$$

$$\text{subject to} \quad (A X - B_2 Y) + (A X - B_2 Y)^T + B_1 B_1^T \preceq 0$$

$$\begin{bmatrix} Z & Y \\ Y^T & X \end{bmatrix} \succeq 0$$

- STRUCTURAL CONSTRAINTS  $F \in \mathcal{S}$

centralized

$$\begin{bmatrix} * & * & * & * \\ * & * & * & * \\ * & * & * & * \\ * & * & * & * \end{bmatrix}$$

fully decentralized

$$\begin{bmatrix} * & & & \\ & * & & \\ & & * & \\ & & & * \end{bmatrix}$$

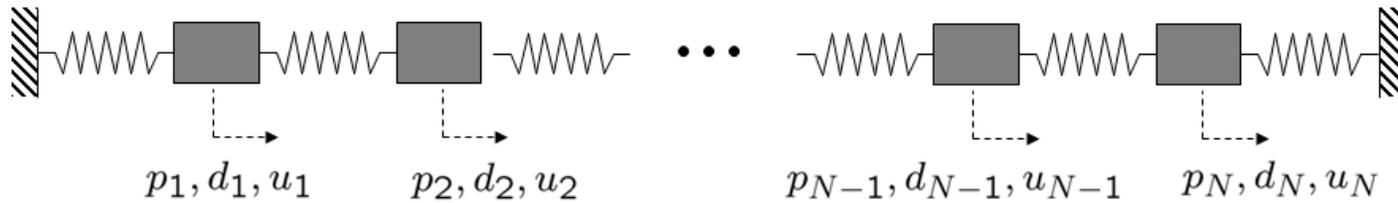
localized

$$\begin{bmatrix} * & * & & \\ * & * & * & \\ & * & * & * \\ & & * & * \end{bmatrix}$$

CHALLENGE:

**convex characterization in the face of structural constraints**

# An example



$$u(t) = - \begin{bmatrix} F_p & F_v \end{bmatrix} \begin{bmatrix} p(t) \\ v(t) \end{bmatrix}$$

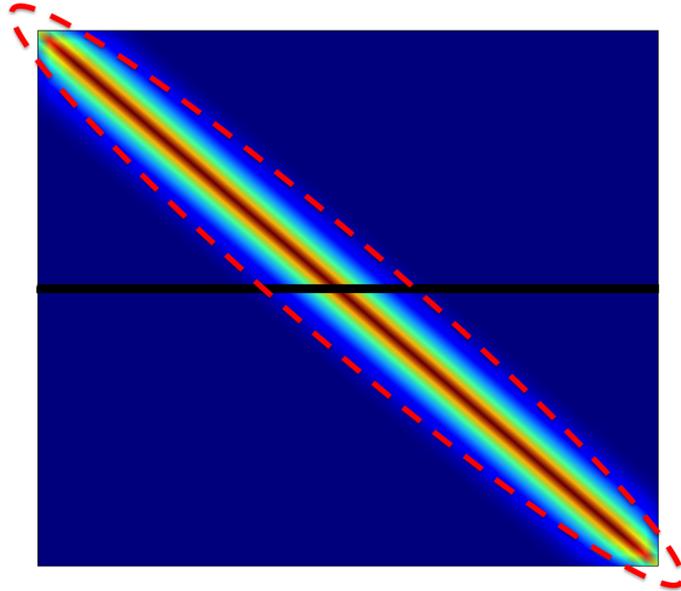
- **Objective: design**  $\begin{bmatrix} F_p & F_v \end{bmatrix}$  **to minimize steady-state variance of**  $p, v, u$

## OPTIMAL CONTROLLER – LINEAR QUADRATIC REGULATOR

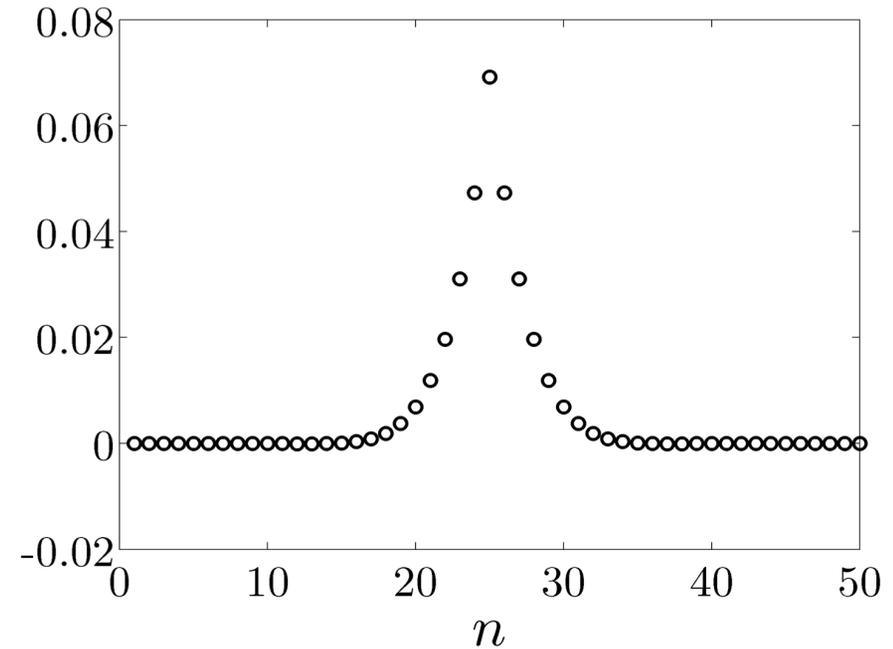
$$\begin{bmatrix} u_1(t) \\ u_2(t) \\ u_3(t) \\ u_4(t) \end{bmatrix} = - \underbrace{\begin{bmatrix} * & * & * & * \\ * & * & * & * \\ * & * & * & * \\ * & * & * & * \end{bmatrix}}_{F_p} \begin{bmatrix} p_1(t) \\ p_2(t) \\ p_3(t) \\ p_4(t) \end{bmatrix} - \underbrace{\begin{bmatrix} * & * & * & * \\ * & * & * & * \\ * & * & * & * \\ * & * & * & * \end{bmatrix}}_{F_v} \begin{bmatrix} v_1(t) \\ v_2(t) \\ v_3(t) \\ v_4(t) \end{bmatrix}$$

# Structure of optimal controller

position feedback matrix:



position gains for middle mass:



## ● OBSERVATIONS

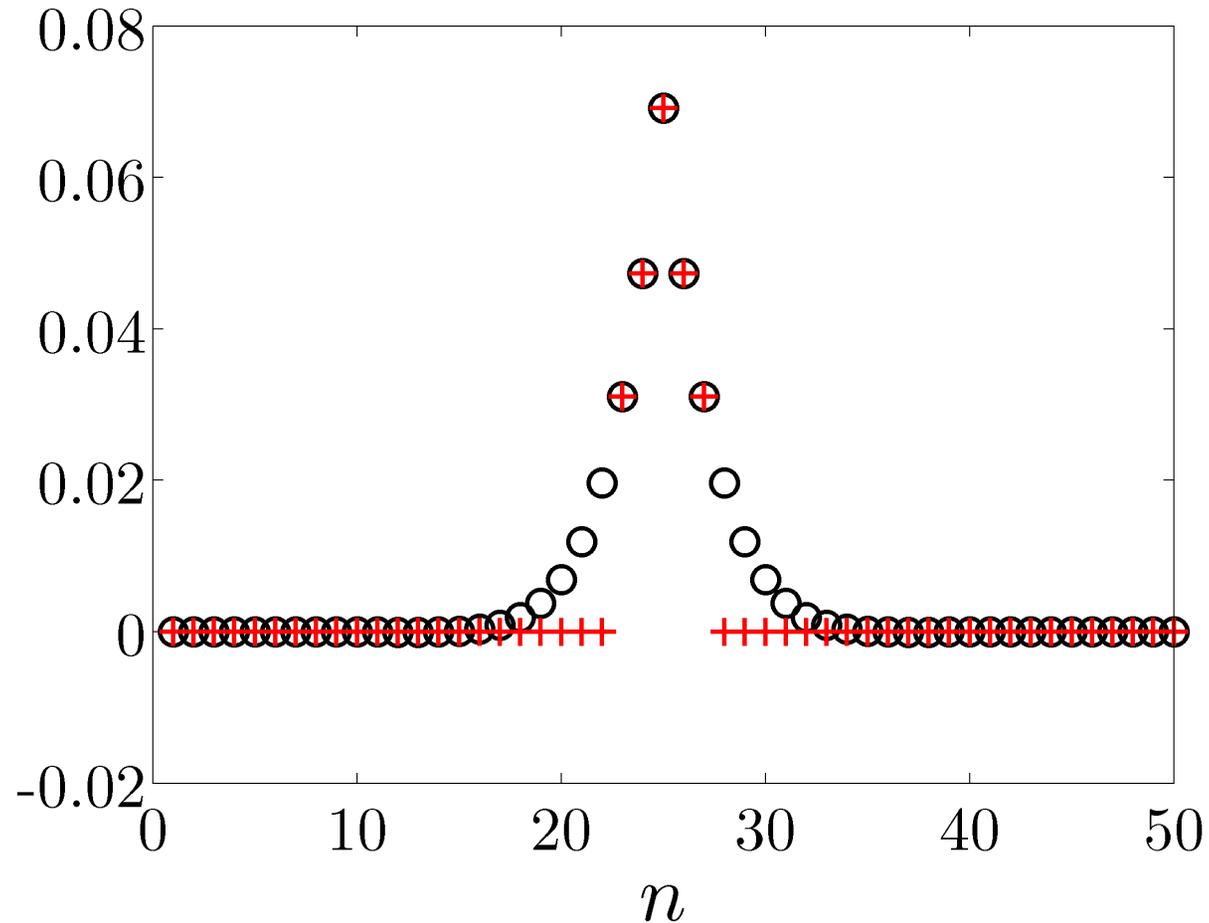
- ★ Diagonals almost constant (modulo edges)
- ★ Off-diagonal decay of a centralized gain

*Bamieh, Paganini, Dahleh, IEEE TAC '02*

*Motee & Jadbabaie, IEEE TAC '08*

## Enforcing localization?

- One approach: **truncating centralized controller**



- POSSIBLE DANGERS

- ★ Performance degradation

- ★ Instability



# SPARSITY-PROMOTING OPTIMAL CONTROL

# Sparsity-promoting optimal control

$$\text{minimize} \quad J(F) \quad + \quad \gamma \mathbf{card}(F)$$

$\downarrow$   
**variance  
amplification**

$\downarrow$   
**sparsity-promoting  
penalty function**

★  $\mathbf{card}(F)$  – number of non-zero elements of  $F$

$$F = \begin{bmatrix} 5.1 & -2.3 & 0 & 1.5 \\ 0 & 3.2 & 1.6 & 0 \\ 0 & -4.3 & 1.8 & 5.2 \end{bmatrix} \Rightarrow \mathbf{card}(F) = 8$$

★  $\gamma > 0$  – performance vs sparsity tradeoff

*Fardad, Lin, Jovanović, ACC '11*

*Lin, Fardad, Jovanović, IEEE TAC '13*

## Convex relaxations of $\text{card}(F)$

$$\ell_1 \text{ norm: } \sum_{i,j} |F_{ij}|$$

$$\text{weighted } \ell_1 \text{ norm: } \sum_{i,j} W_{ij} |F_{ij}|, \quad W_{ij} \geq 0$$

- CARDINALITY VS WEIGHTED  $\ell_1$  NORM

$$\{W_{ij} = 1/|F_{ij}|, F_{ij} \neq 0\} \Rightarrow \text{card}(F) = \sum_{i,j} W_{ij} |F_{ij}|$$

### RE-WEIGHTED SCHEME

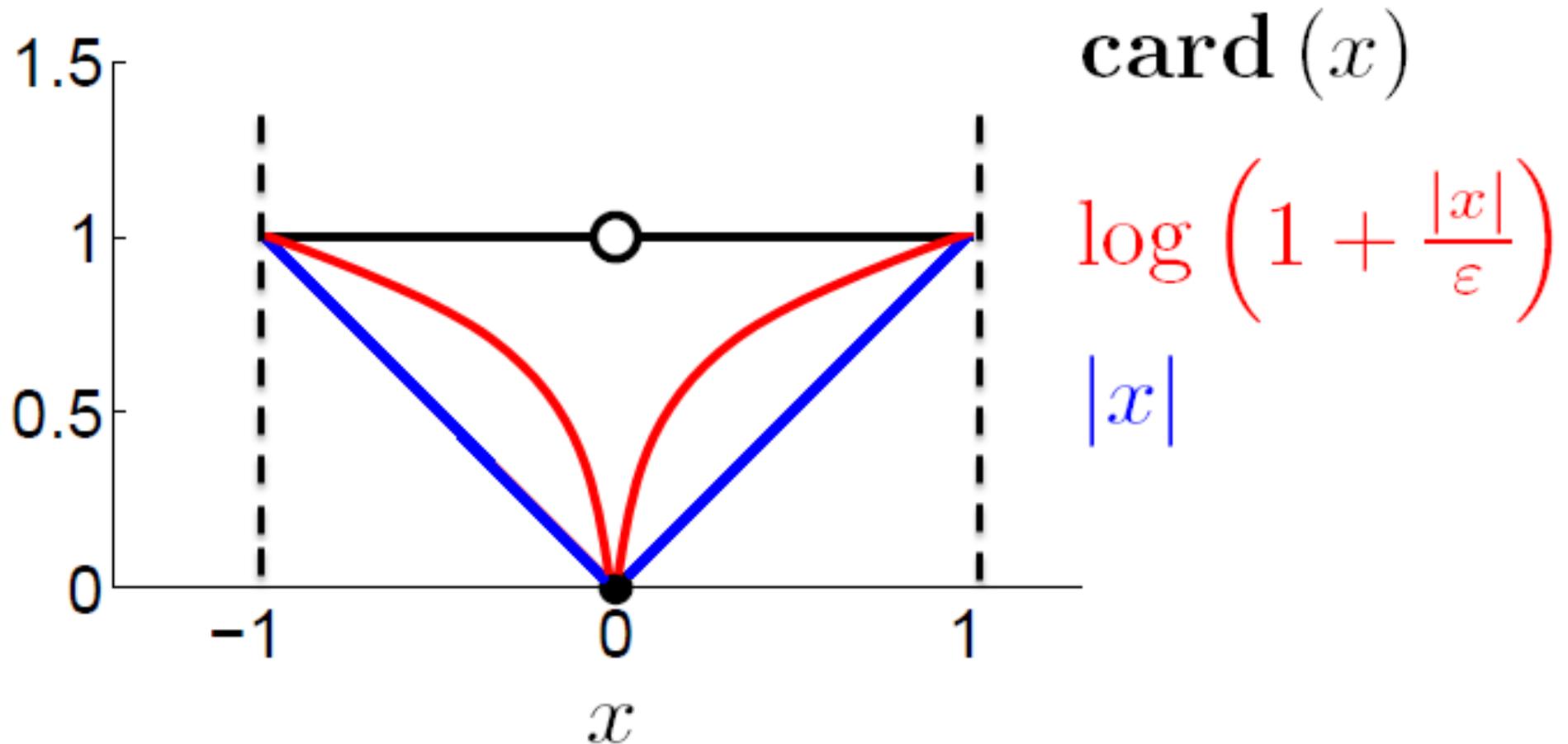
- ★ Use feedback gains from previous iteration to form weights

$$W_{ij}^+ = \frac{1}{|F_{ij}| + \varepsilon}$$

*Candès, Wakin, Boyd, J. Fourier Anal. Appl. '08*

## A non-convex relaxation of card ( $F$ )

**sum-of-logs:**  $\sum_{i,j} \log \left( 1 + \frac{|F_{ij}|}{\varepsilon} \right), \quad 0 < \varepsilon \ll 1$



# CLASSES OF CONVEX PROBLEMS

# Optimal actuator/sensor selection

- OBJECTIVE: identify **row-sparse** feedback gain

$$u = -Fx$$

minimize

$$J(F)$$

+

$$\gamma \sum_i \|e_i^T F\|_2$$



**variance  
amplification**



**row-sparsity-promoting  
penalty function**

- CHANGE OF VARIABLES:  $Y := F X$ 
  - ★ **convex dependence** of  $J$  on  $X$  and  $Y$
  - ★ **row-sparse structure preserved**

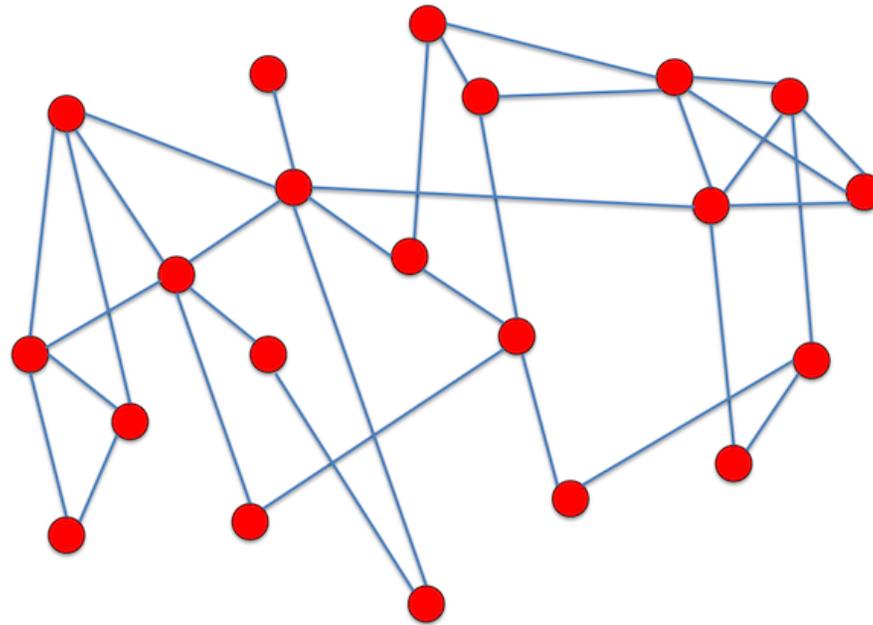
$$u = - F x = - Y X^{-1} x$$

- OPTIMAL CONTROL PROBLEM
  - ★ **admits SDP characterization**

*Polyak, Khlebnikov, Shcherbakov, ECC '13*

*Dhingra, Jovanović, Luo, CDC '14*

# Consensus by distributed computation



- RELATIVE INFORMATION EXCHANGE WITH NEIGHBORS
  - ★ simplest **distributed averaging** algorithm

$$\dot{x}_i(t) = - \sum_{j \in \mathcal{N}_i} (x_i(t) - x_j(t))$$

**connected network**  $\Rightarrow$  **convergence to the average value**

## Consensus with stochastic disturbances

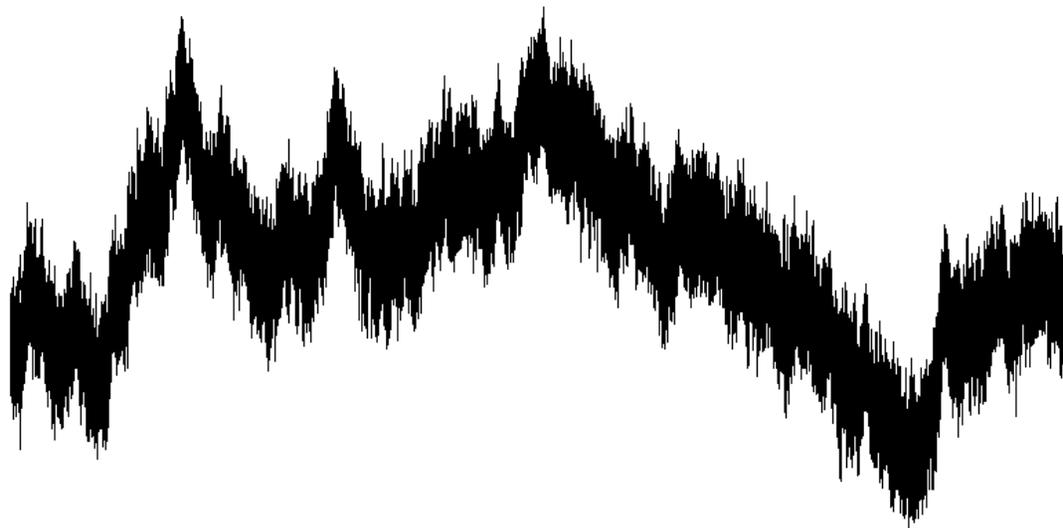
$$\dot{x}_i(t) = - \sum_{j \in \mathcal{N}_i} (x_i(t) - x_j(t)) + d_i(t)$$



**white noise**

- NETWORK AVERAGE

- ★ **undergoes random walk**



connected network  $\Rightarrow$   $\left\{ \begin{array}{l} \text{each } x_i(t) \text{ fluctuates around } \bar{x}(t) \\ \text{deviation from average: } \tilde{x}_i(t) := x_i(t) - \bar{x}(t) \end{array} \right.$

## Design of undirected consensus networks

dynamics:  $\dot{x} = d + u$

control:  $u = -F x$

objective:  $J = \lim_{t \rightarrow \infty} \mathcal{E} (x^T(t) Q x(t) + u^T(t) R u(t))$  ■

### SDP characterization:

minimize  $\text{trace}(X + R F) + \gamma \mathbf{1}^T Y \mathbf{1}$

subject to  $\begin{bmatrix} X & Q^{1/2} \\ Q^{1/2} & F + \mathbf{1}\mathbf{1}^T/N \end{bmatrix} \succeq 0$

$$-Y_{ij} \leq W_{ij} F_{ij} \leq Y_{ij}$$

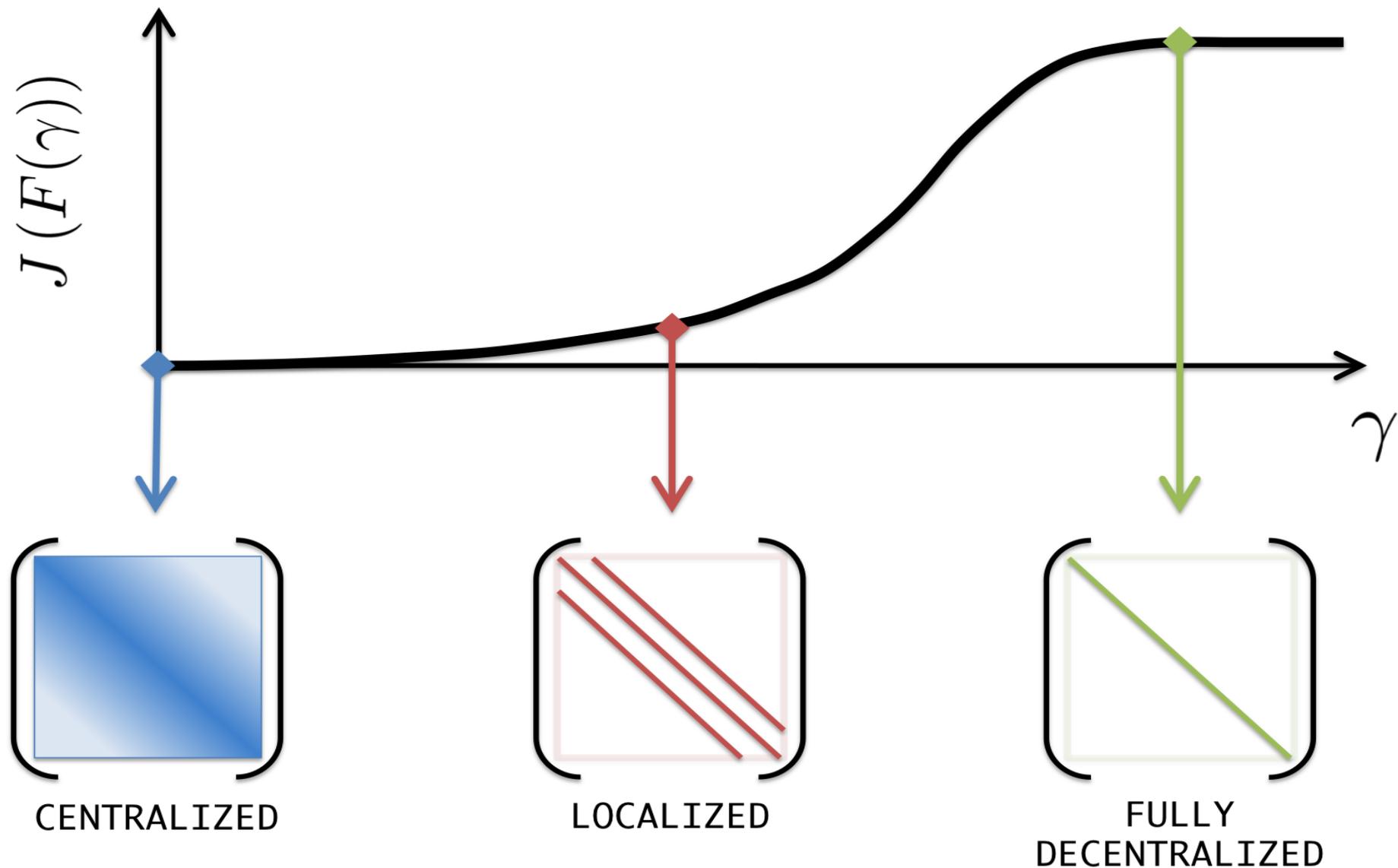
$$F \mathbf{1} = 0$$

*Lin, Fardad, Jovanović, Allerton '12*

*Wu & Jovanović, ACC '14*

## Parameterized family of feedback gains

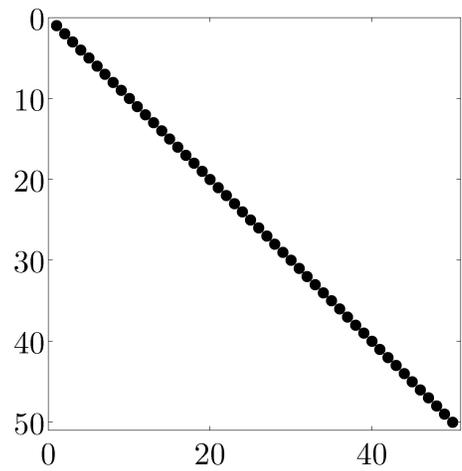
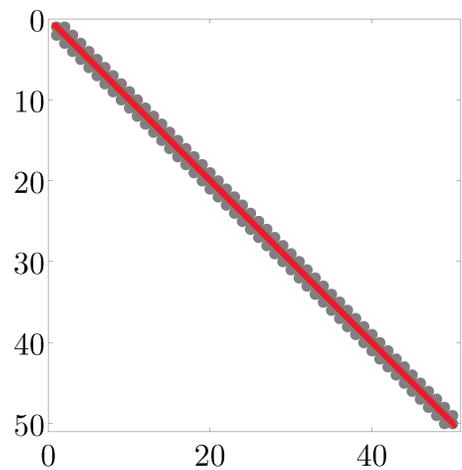
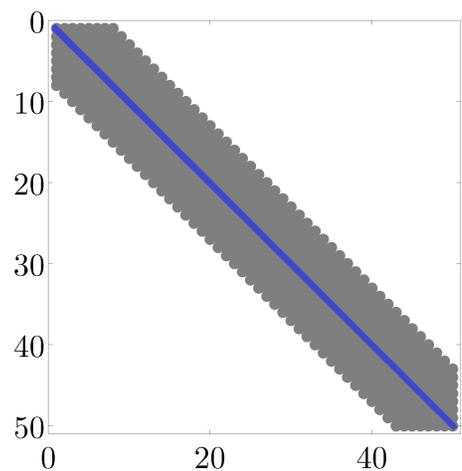
$$F(\gamma) := \operatorname{argmin}_F (J(F) + \gamma g(F))$$



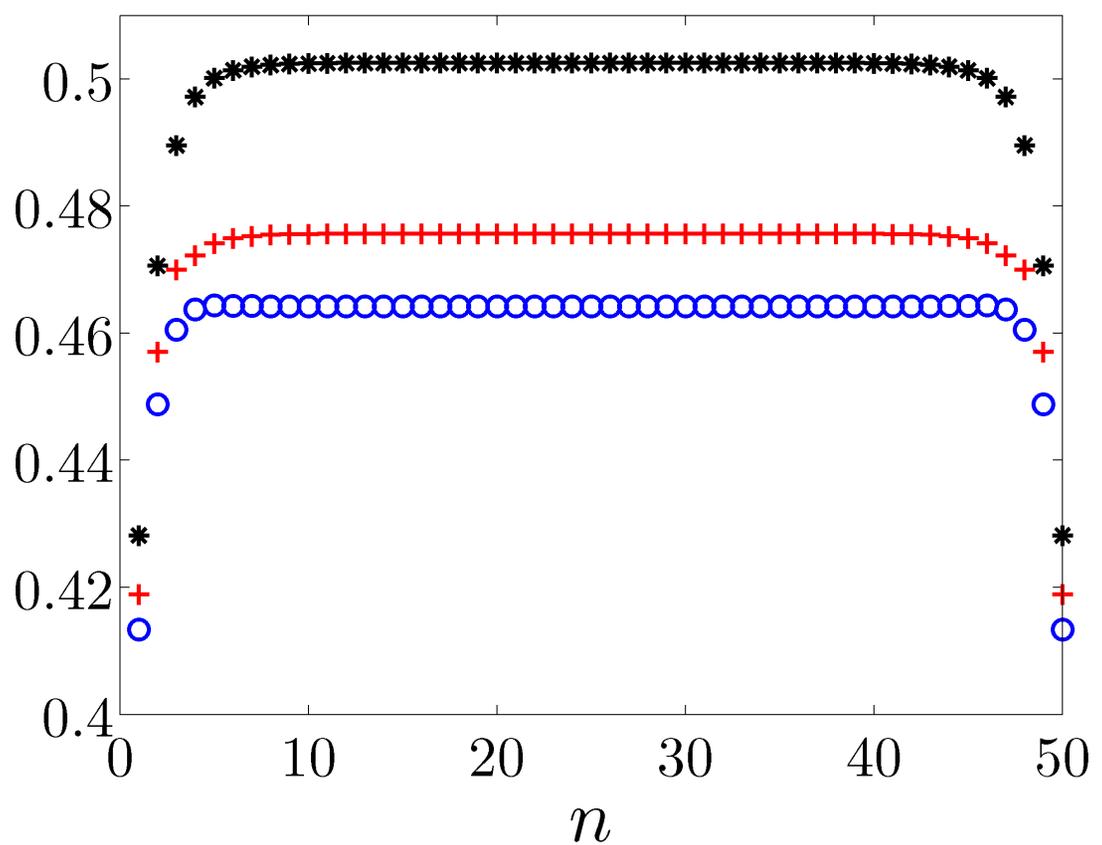
# EXAMPLES

[www.umn.edu/~mihailo/software/lqrsp/](http://www.umn.edu/~mihailo/software/lqrsp/)

# Mass-spring system



diag( $F_v$ ):

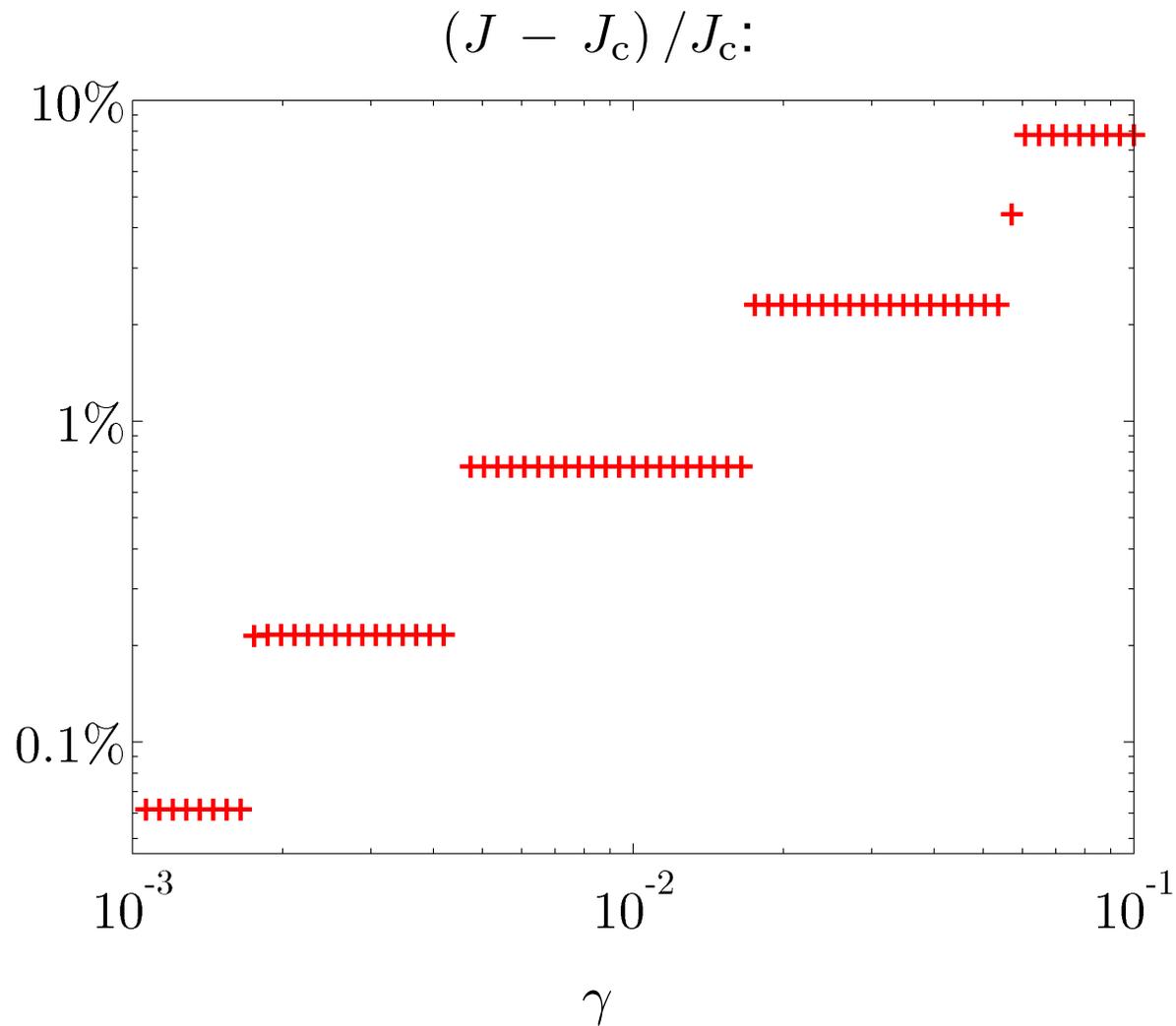


$\gamma = 10^{-4}$

$\gamma = 0.03$

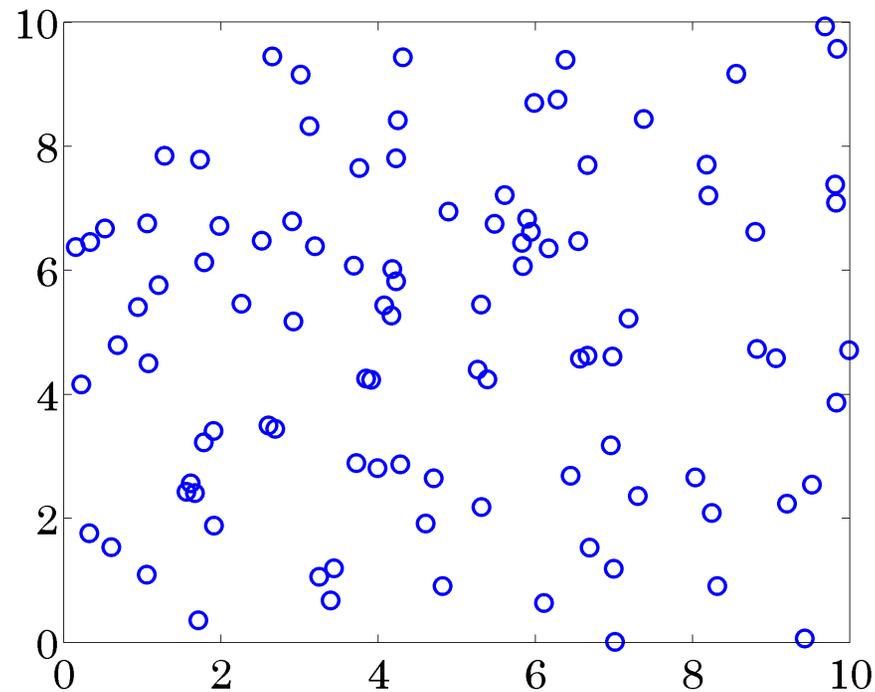
$\gamma = 0.1$

- Performance comparison: **sparse vs centralized**



$\text{card}(F) / \text{card}(F_c)$	$(J - J_c) / J_c$
10%	0.75%
6%	2.4%
2%	7.8%

## Network with 100 nodes

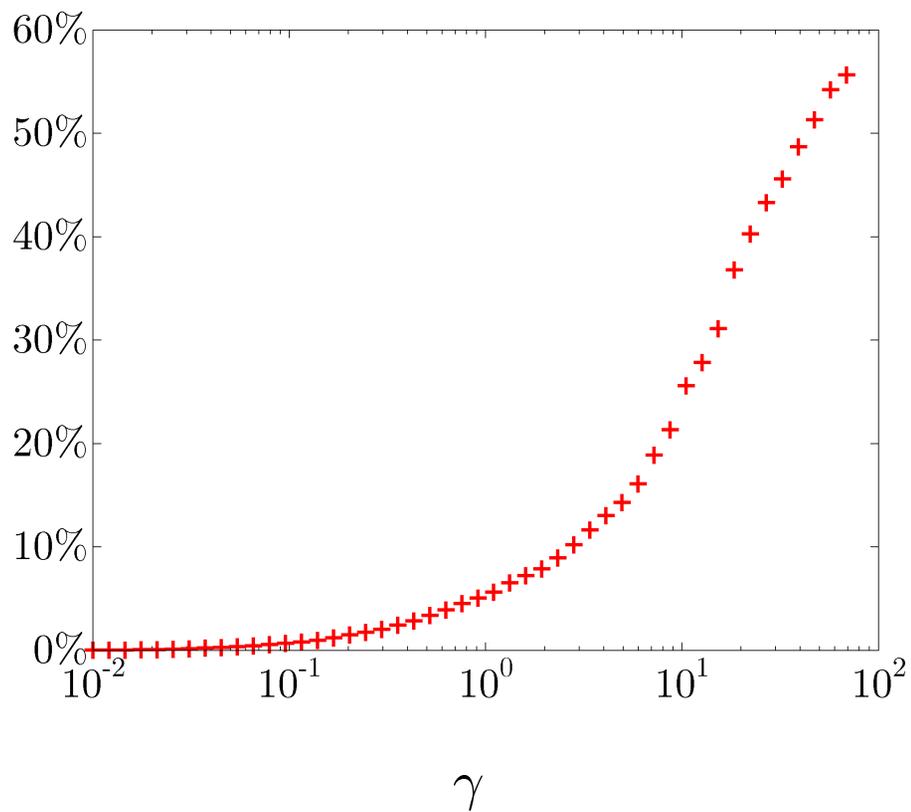


$$\begin{bmatrix} \dot{p}_i \\ \dot{v}_i \end{bmatrix} = \underbrace{\begin{bmatrix} 1 & 1 \\ 1 & 2 \end{bmatrix} \begin{bmatrix} p_i \\ v_i \end{bmatrix}}_{\text{unstable dynamics}} + \underbrace{\sum_{j \neq i} e^{-\alpha(i,j)} \begin{bmatrix} p_j \\ v_j \end{bmatrix}}_{\text{coupling}} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} (d_i + u_i)$$

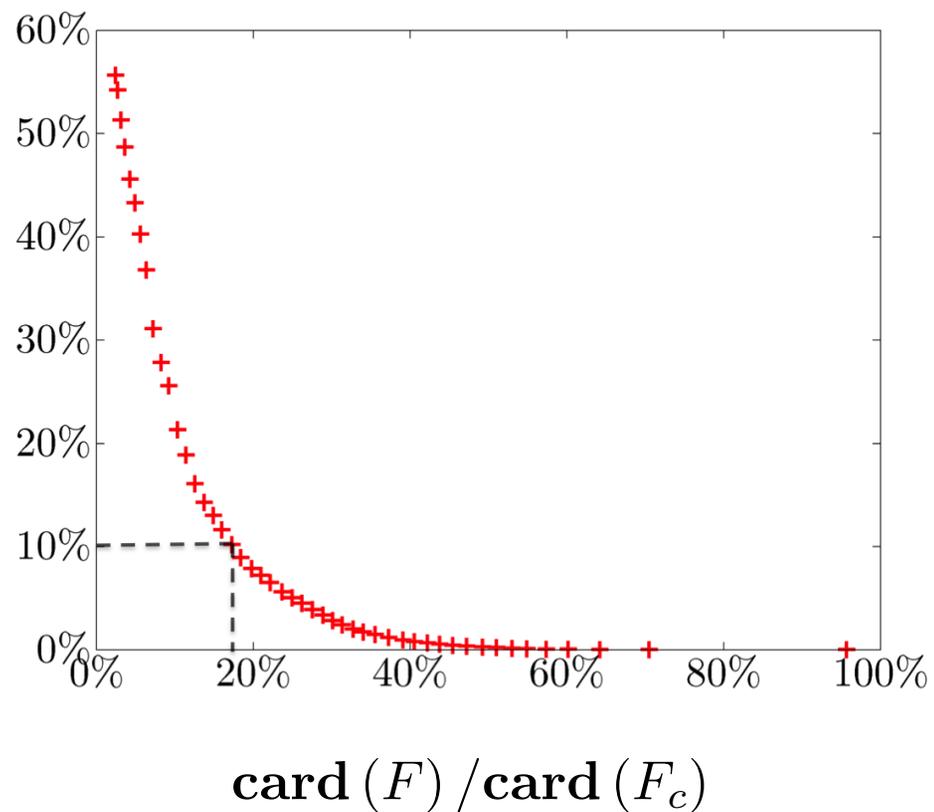
$\alpha(i, j)$ : Euclidean distance between nodes  $i$  and  $j$

- Performance comparison: **sparse vs centralized**

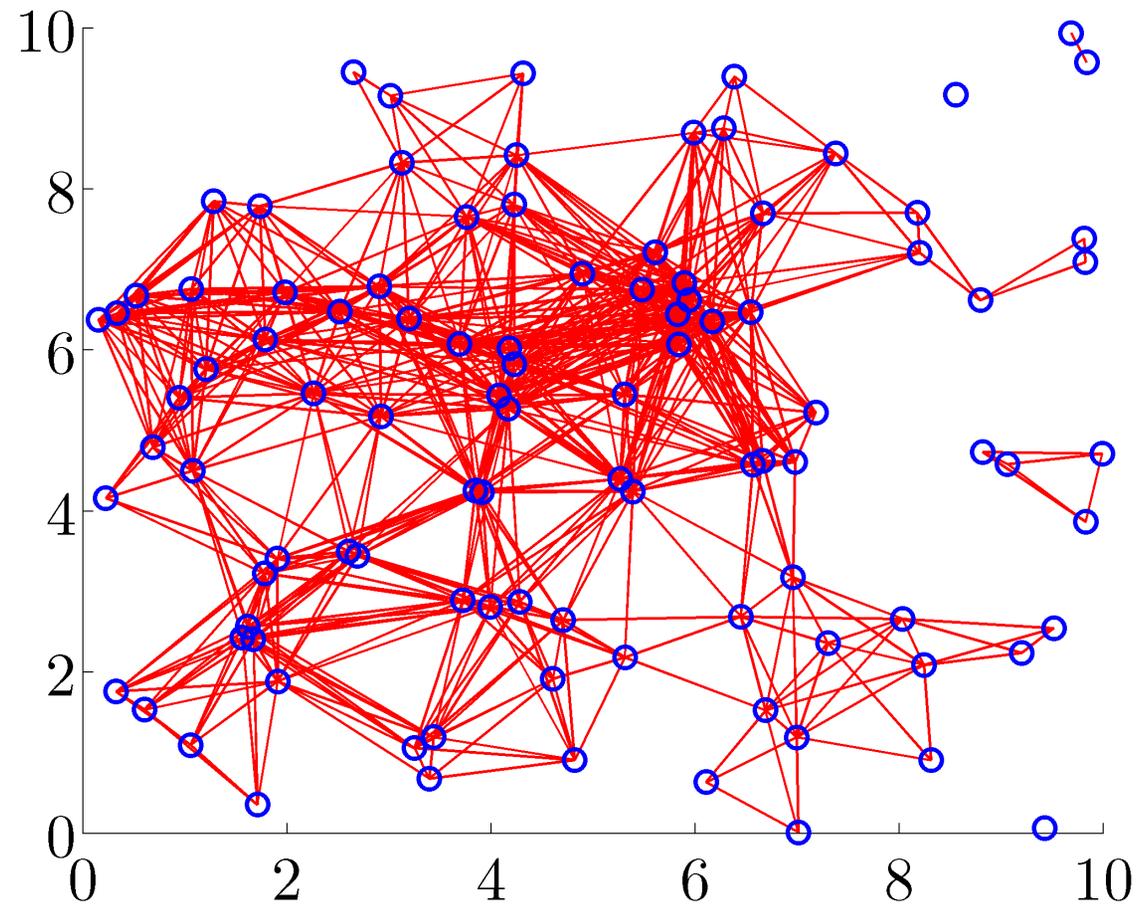
$$(J - J_c) / J_c:$$



$$(J - J_c) / J_c:$$



## identified communication graph:

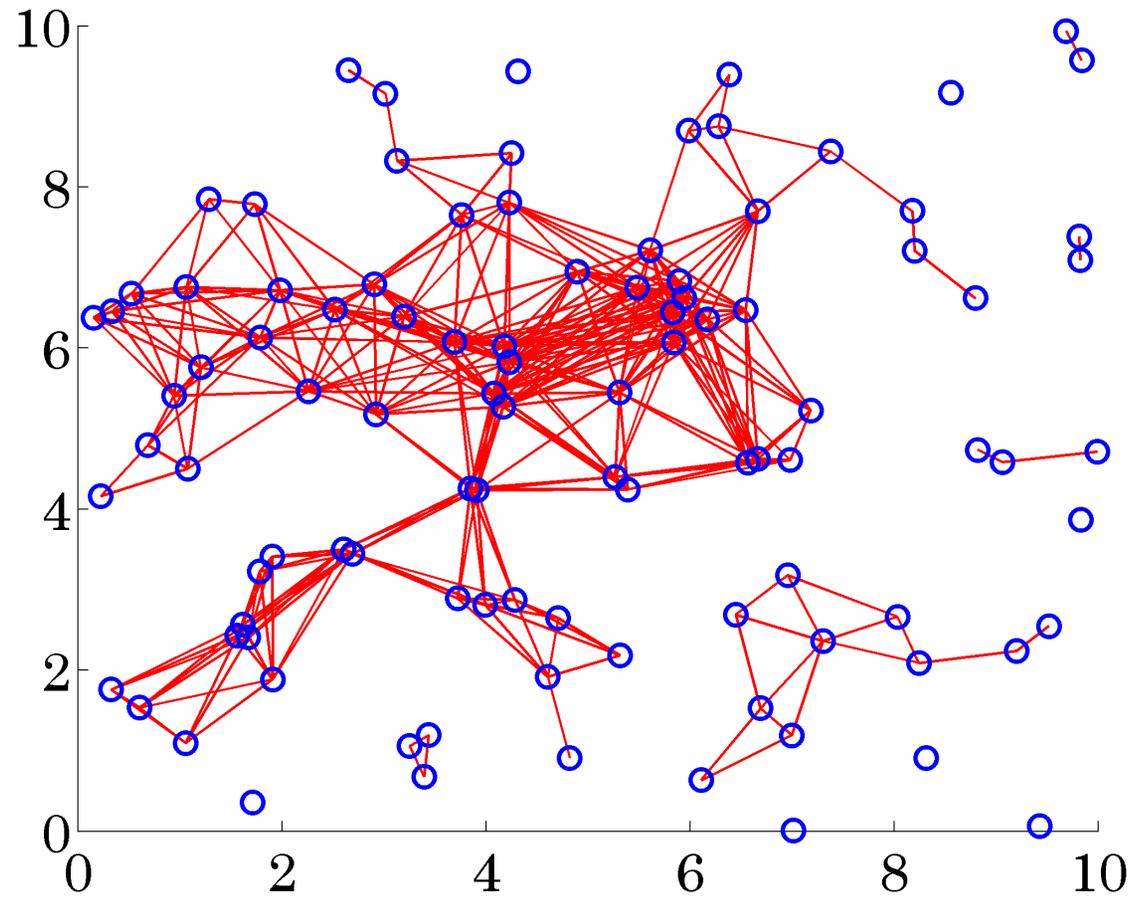


$$\gamma = 5$$

$$\text{card}(F) / \text{card}(F_c) = 8.8\%$$

$$(J - J_c) / J_c = 24.6\%$$

## identified communication graph:

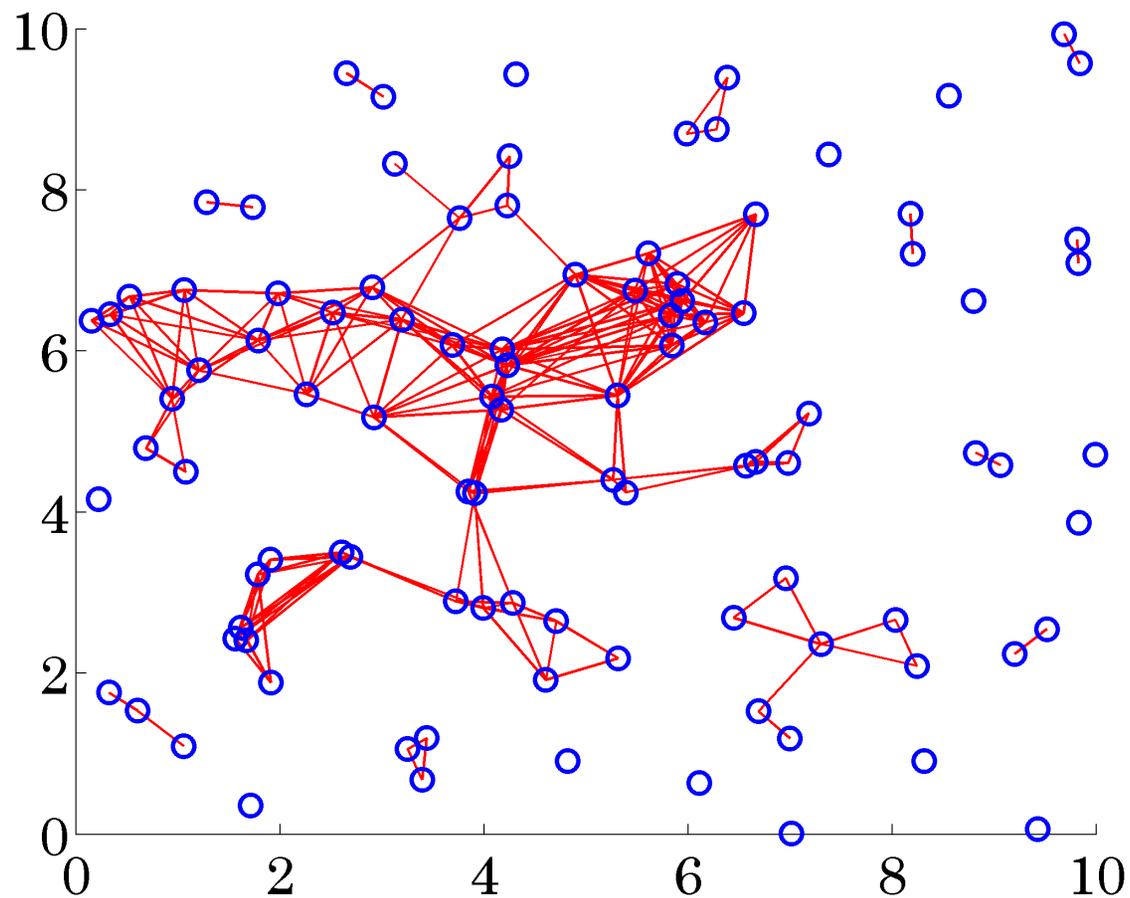


$$\gamma = 11$$

$$\text{card}(F) / \text{card}(F_c) = 5.1\%$$

$$(J - J_c) / J_c = 40.9\%$$

## identified communication graph:

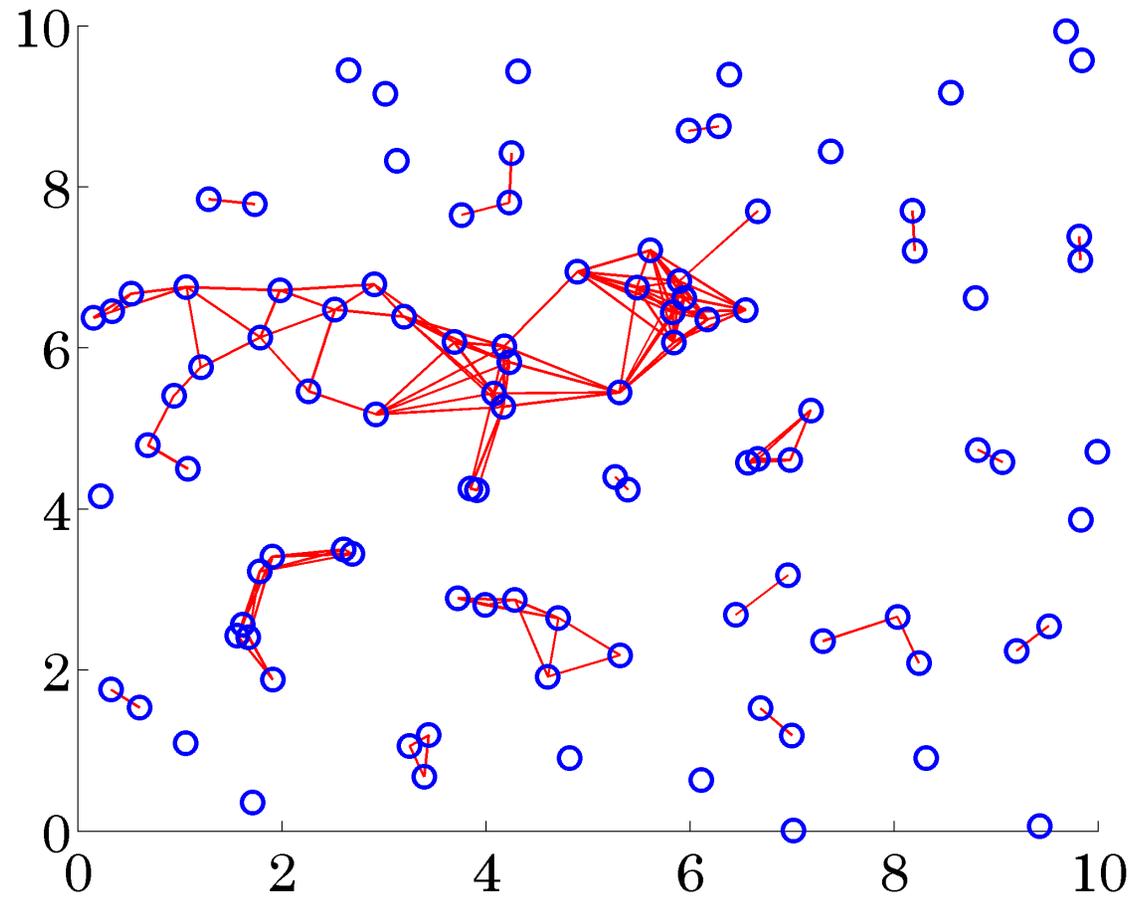


$$\gamma = 18$$

$$\text{card}(F) / \text{card}(F_c) = 3.4\%$$

$$(J - J_c) / J_c = 48.7\%$$

## identified communication graph:

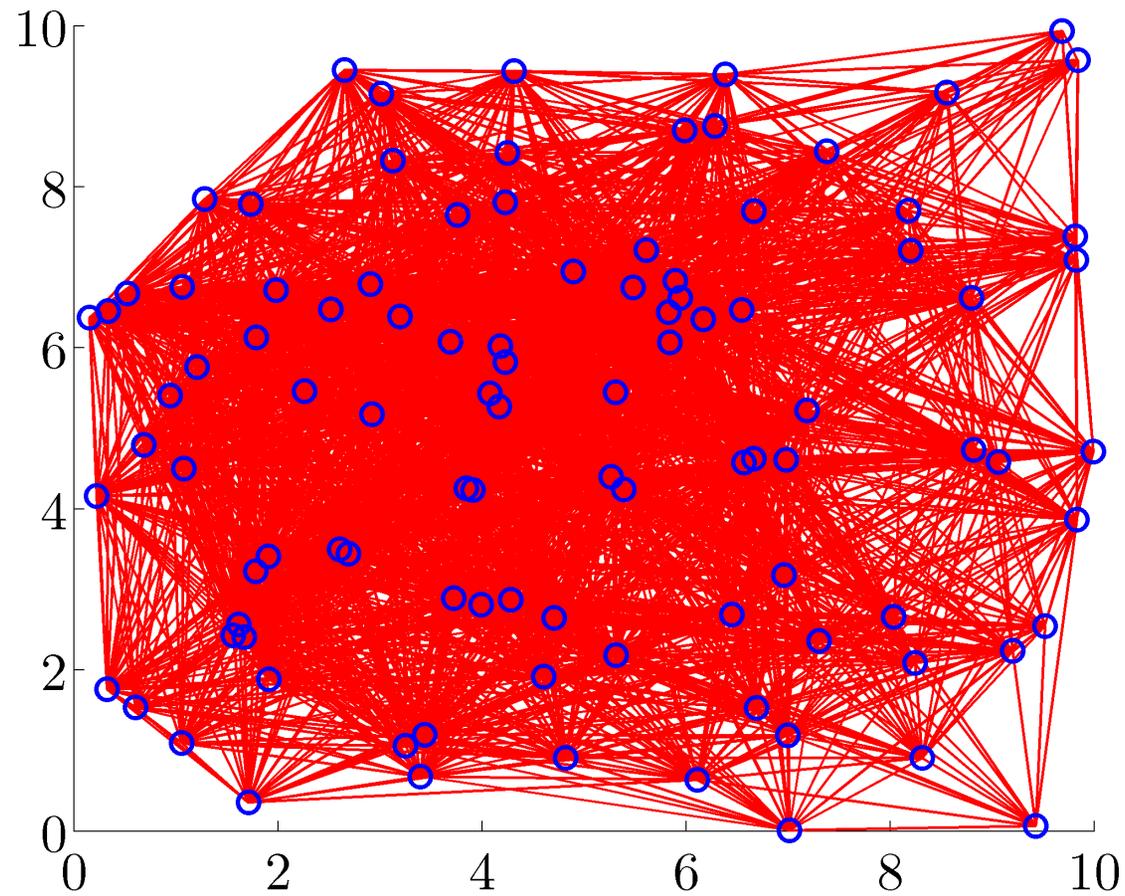


$$\gamma = 30$$

$$\text{card}(F) / \text{card}(F_c) = 2.4\%$$

$$(J - J_c) / J_c = 54.8\%$$

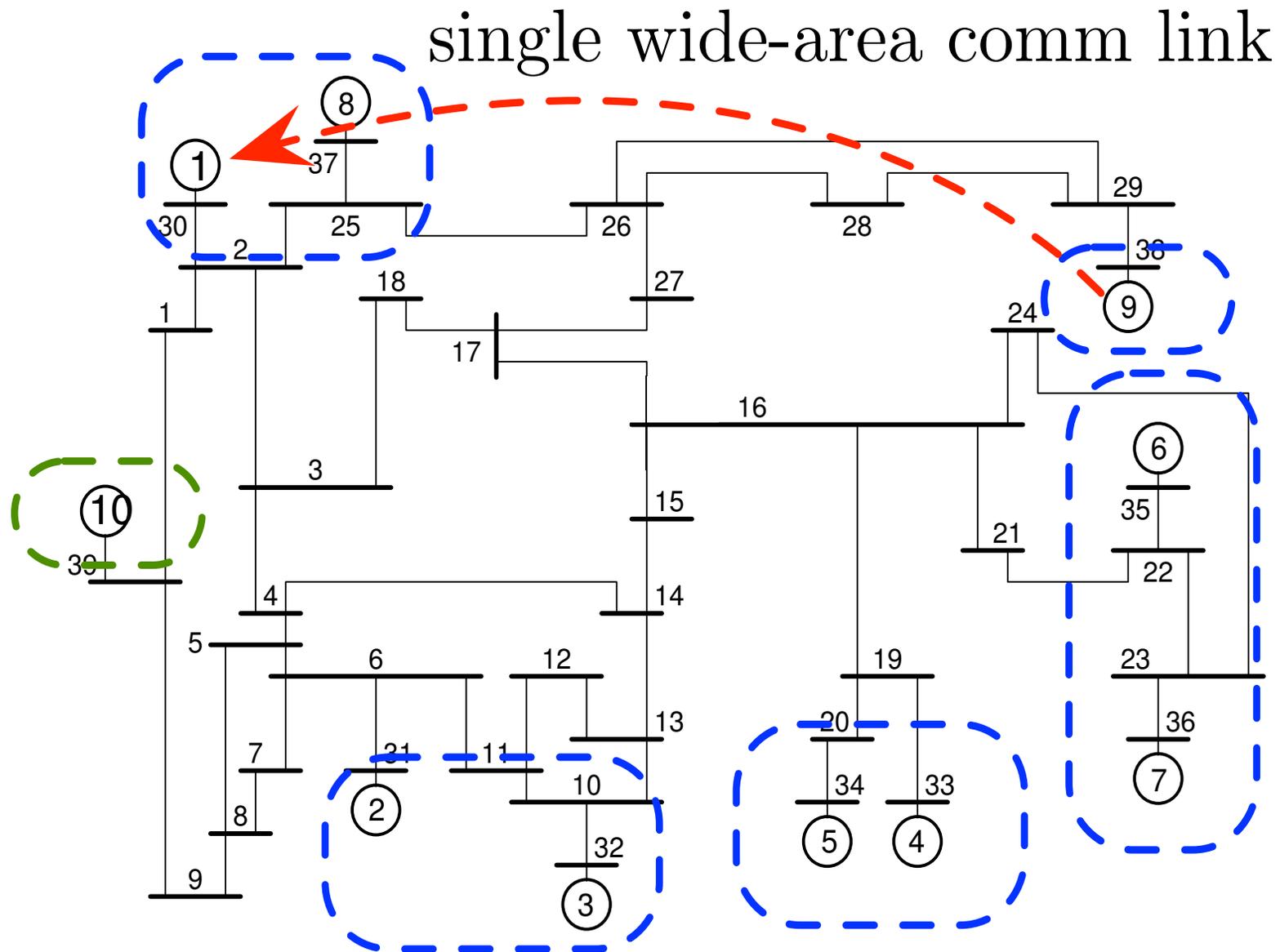
communication graph of a truncated centralized gain:



$$\text{card}(F) = 7380 \text{ (36.9\%)}$$

**non-stabilizing**

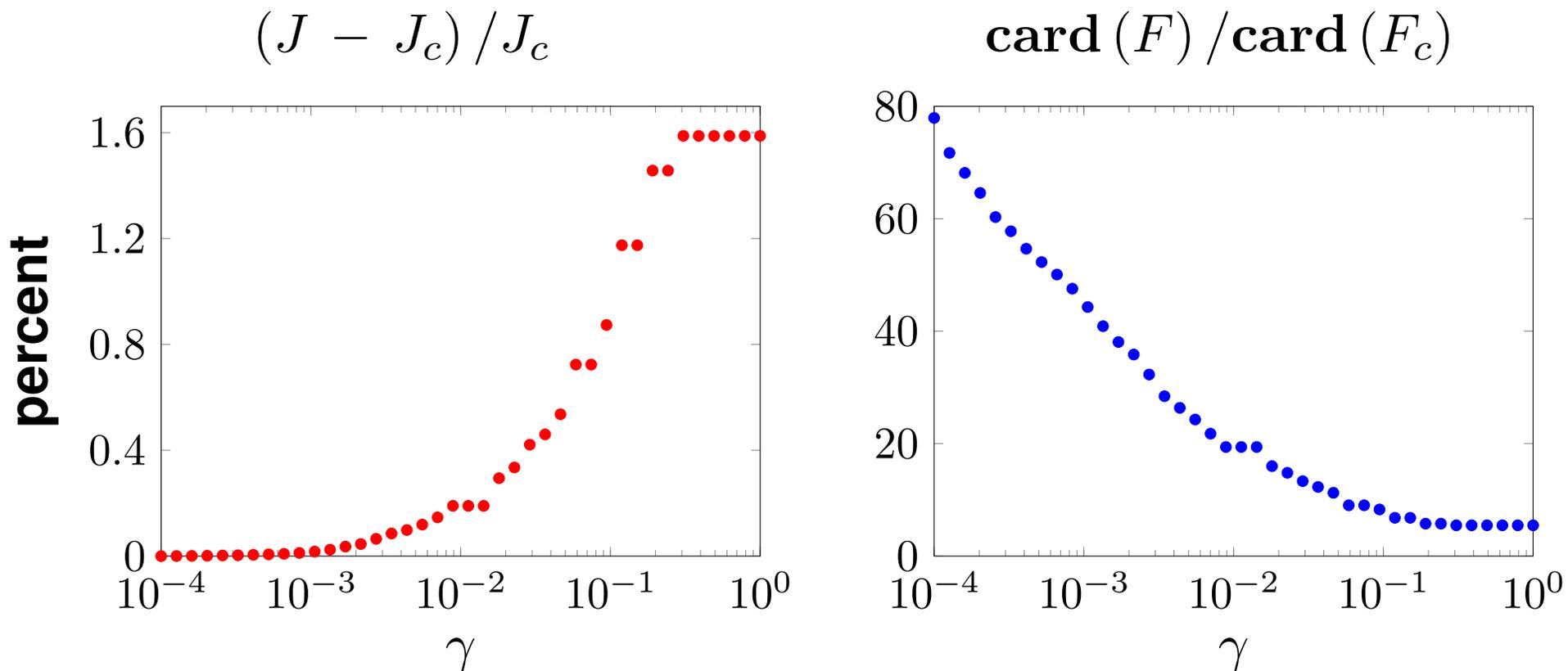
# Wide area control of power networks



single long range interaction  $\Rightarrow$

**nearly centralized performance**

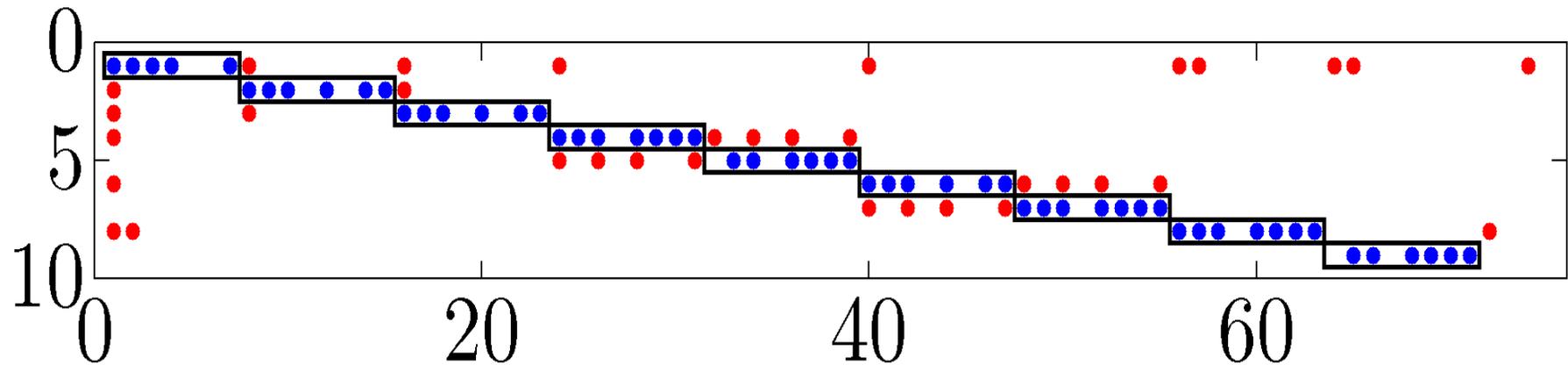
## Performance vs sparsity



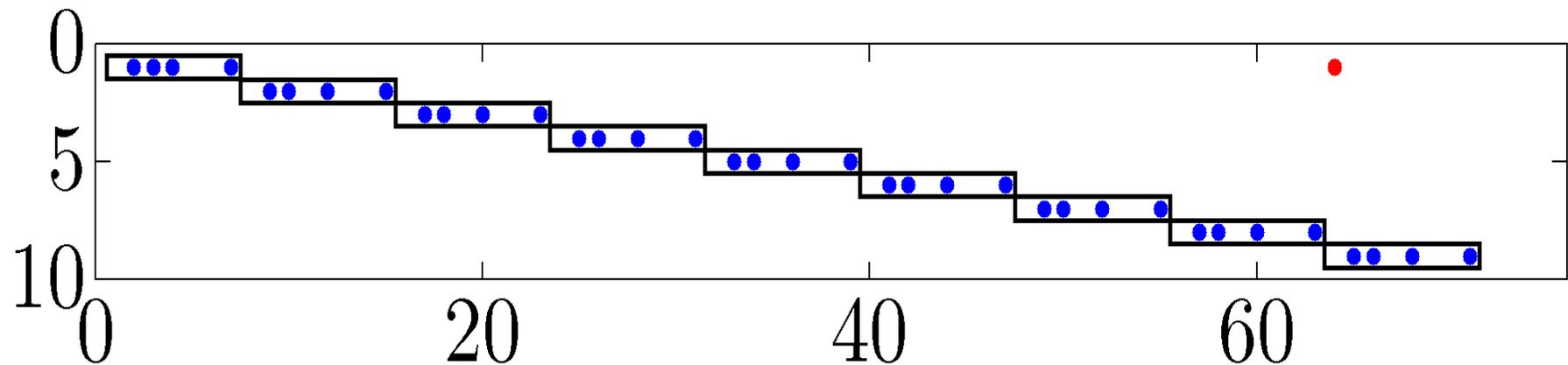
$$\gamma = 1 \xrightarrow{\text{relative to } F_c} \begin{cases} 1.6 \% & \text{performance loss} \\ 5.5 \% & \text{non-zero elements in } F \end{cases}$$

- Signal exchange network

$$\gamma = 0.0289, \text{card}(F) = 90$$

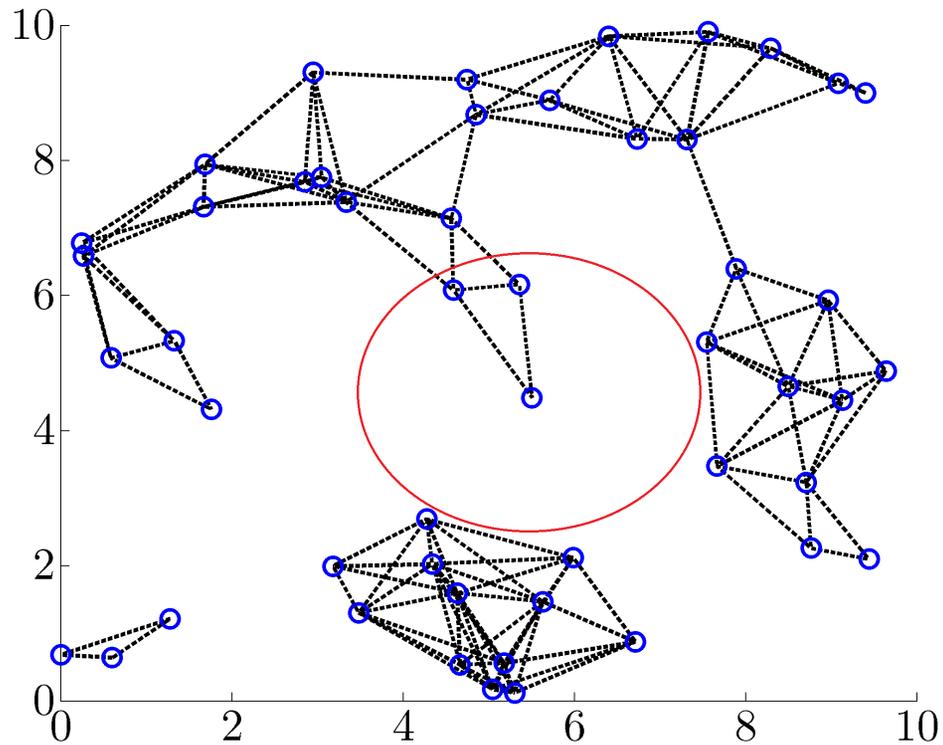


$$\gamma = 1, \text{card}(F) = 37$$



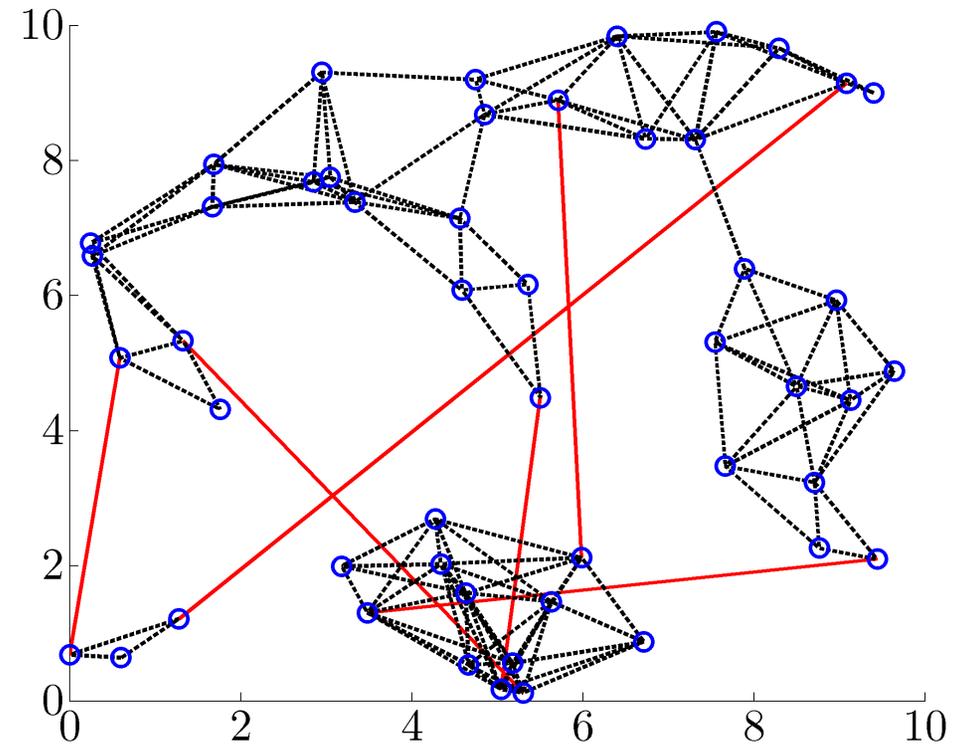
# Sparsity-promoting consensus algorithm

**local performance graph:**



$$Q = Q_{\text{loc}} + \left( I - \frac{1}{N} \mathbf{1}\mathbf{1}^T \right)$$

**identified communication graph:**



$$\frac{J - J_c}{J_c} \approx 11\%$$

*Lin, Fardad, Jovanović, Allerton '12*

# ALGORITHM

# Alternating direction method of multipliers

$$\text{minimize } J(F) + \gamma g(F)$$

- **Step 1: introduce additional variable/constraint**

$$\begin{array}{l} \text{minimize } J(F) + \gamma g(G) \\ \text{subject to } F - G = 0 \end{array}$$

**benefit: decouples  $J$  and  $g$**

- **Step 2: introduce augmented Lagrangian**

$$\mathcal{L}_\rho(F, G, \Lambda) = J(F) + \gamma g(G) + \text{trace}(\Lambda^T(F - G)) + \frac{\rho}{2} \|F - G\|_F^2$$

- **Step 3: use ADMM for augmented Lagrangian minimization**

$$\mathcal{L}_\rho(F, G, \Lambda) = J(F) + \gamma g(G) + \text{trace}(\Lambda^T(F - G)) + \frac{\rho}{2} \|F - G\|_F^2$$

ADMM:

$$F^{k+1} := \underset{F}{\operatorname{argmin}} \mathcal{L}_\rho(F, G^k, \Lambda^k)$$

$$G^{k+1} := \underset{G}{\operatorname{argmin}} \mathcal{L}_\rho(F^{k+1}, G, \Lambda^k)$$

$$\Lambda^{k+1} := \Lambda^k + \rho(F^{k+1} - G^{k+1})$$

### MANY MODERN APPLICATIONS

- ★ distributed computing
- ★ distributed signal processing
- ★ image denoising
- ★ machine learning

- **Step 4: Polishing** – back to structured optimal design

★ ADMM  $\left\{ \begin{array}{l} \text{identifies sparsity patterns} \\ \text{provides good initial condition for structured design} \end{array} \right.$

★ NECESSARY CONDITIONS FOR OPTIMALITY OF THE STRUCTURED PROBLEM

$$\begin{aligned} (A - B_2 F)^T P + P (A - B_2 F) &= -(Q + F^T R F) \\ (A - B_2 F) L + L (A - B_2 F)^T &= -B_1 B_1^T \\ [(R F - B_2^T P) L] \circ I_S &= 0 \end{aligned}$$

**Newton's method with conjugate gradient**

$I_S$  - structural identity

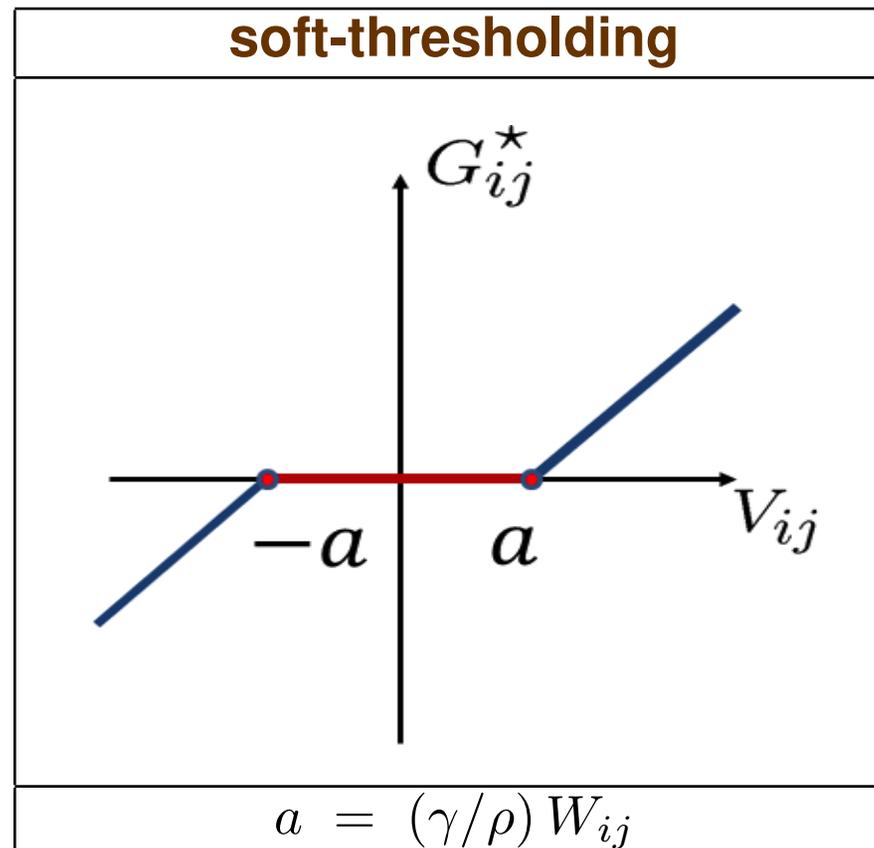
$$F = \begin{bmatrix} * & * & & & \\ * & * & * & & \\ & * & * & * & \\ & & * & * & \end{bmatrix} \Rightarrow I_S = \begin{bmatrix} 1 & 1 & & & \\ 1 & 1 & 1 & & \\ & 1 & 1 & 1 & \\ & & 1 & 1 & \end{bmatrix}$$

## Solution to $G$ -minimization problem

$$\underset{G_{ij}}{\text{minimize}} \sum_{i,j} \left( \gamma W_{ij} |G_{ij}| + \frac{\rho}{2} (G_{ij} - V_{ij}^k)^2 \right)$$

$$V_{ij}^k := F_{ij}^{k+1} + (1/\rho)\Lambda_{ij}^k$$

**separability**  $\Rightarrow$  **element-wise analytical solution**



## Solution to $F$ -minimization problem

$$\underset{F}{\text{minimize}} \quad J(F) + \frac{\rho}{2} \|F - U^k\|_F^2$$

$$U^k := G^k - (1/\rho)\Lambda^k$$

NECESSARY CONDITIONS FOR OPTIMALITY:

$$\begin{aligned} (A - B_2 F)L + L(A - B_2 F)^T &= -B_1 B_1^T \\ (A - B_2 F)^T P + P(A - B_2 F) &= -(Q + F^T R F) \\ FL + \rho(2R)^{-1}F &= R^{-1}B_2^T PL + \rho(2R)^{-1}U^k \end{aligned}$$

- ITERATIVE SCHEME

Given  $F_0$  solve for  $\{L_1, P_1\} \rightarrow F_1 \rightarrow \{L_2, P_2\} \rightarrow F_2 \dots$

**descent direction** + **line search**  $\Rightarrow$  **convergence**

# ALTERNATIVE FORMULATIONS?

# Optimal control in discrete time

$$x_{t+1} = A x_t + B_1 d_t + B_2 u_t$$

$$u_t = -F x_t$$

- NO STRUCTURAL CONSTRAINTS

$$\underset{X, F}{\text{minimize}} \quad \text{trace} (X B_1 B_1^T)$$

$$\text{subject to} \quad X - (A - B_2 F)^T X (A - B_2 F) = Q + F^T R F$$

$$X \succ 0$$

## equivalent formulation:

$$\underset{X, Y, F, K}{\text{minimize}} \quad \text{trace} (X B_1 B_1^T)$$

$$\text{subject to} \quad X - (A - B_2 F)^T Y^{-1} (A - B_2 F) \succeq Q + K$$

$$X \succ 0, \quad K \succeq F^T R F$$

$$XY = I$$

- A POSSIBLE APPROACH

$$\underset{X, Y, F, K}{\text{minimize}} \quad \text{trace} (X B_1 B_1^T)$$

$$\text{subject to} \quad X - (A - B_2 F)^T Y^{-1} (A - B_2 F) \succeq Q + K$$

$$K \succeq F^T R F, \quad X \succeq Y^{-1}, \quad Y \preceq X^{-1}$$

$$X \succ 0, \quad Y \succ 0$$

## CONVEX APPROXIMATION?

$$\underset{X, Y, F, K}{\text{minimize}} \quad \text{trace} (X Y_k + X_k Y) + \text{trace} (X B_1 B_1^T)$$

$$\text{subject to} \quad \begin{bmatrix} X - Q - K & (A - B_2 F)^T \\ A - B_2 F & Y \end{bmatrix} \succeq 0$$

$$\begin{bmatrix} K & F^T \\ F & R^{-1} \end{bmatrix} \succeq 0, \quad \begin{bmatrix} X & I \\ I & Y \end{bmatrix} \succeq 0$$

# Summary

- SPARSITY-PROMOTING OPTIMAL CONTROL

- ★ Performance vs sparsity tradeoff

*Lin, Fardad, Jovanović, IEEE TAC '13*

- ★ Software

[www.umn.edu/~mihailo/software/lqrsp/](http://www.umn.edu/~mihailo/software/lqrsp/)

- RELATED EFFORT

- ★ Leader selection in large dynamic networks

*Lin, Fardad, Jovanović, IEEE TAC '14*

- ★ Optimal synchronization of sparse oscillator networks

*Fardad, Lin, Jovanović, IEEE TAC '14*

- ★ Optimal dissemination of information in social networks

*Fardad, Zhang, Lin, Jovanović, CDC '12*

- ★ Sparse or infrequently changing (in time) control signals

*Jovanović & Lin, ECC '13*

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## DISCUSSIONS:

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