

Application of Symbolic Analysis to Power and Ground Interconnect Optimization



Sachin Sapatnekar
Department of Electrical
and Computer Engineering
Iowa State University
Ames, Iowa 50011, USA
Tel: (515) 294-1426
Fax: (515) 294-8432
sachin@iastate.edu

Jatan Shah
MIPS Technology Inc.
2011 N. Shoreline Blvd
Mountain View
California 94043, USA

Marwan Hassoun
Department of Electrical
and Computer Engineering
Iowa State University
Ames, Iowa 50011, USA
Tel: (515) 294-2663
Fax: (515) 294-8432
marwan@iastate.edu

OUTLINE

- **Motivation**
- **Specific Power and Ground Problems and Solutions**
- **Symbolic Analysis Role**
- **Modeling of Power and Ground Buses**
- **Admittance and Current Estimation Techniques**
- **Moment Computations**
- **Time Domain Transformation**
- **Preliminary Experimental Results**
- **Conclusions**
- **Future Work**

MOTIVATION

- **High-speed circuits ->Gigahertz clock rates -> Increased switching activity**
- **Feature sizes to deep submicron levels**
- **Vdd levels going down to conserve power**
- **Power and Ground buses cannot be assumed to be perfect conductors**
- **P&G buses must be designed carefully to ensure that supply voltage levels are maintained at appropriate levels.**
- **Better CAD techniques needed for estimation of voltage drops in P&G buses for functional and physical reliability in integrated circuits.**

SPECIFIC P&G PROBLEMS

- **Functional and Physical reliability**
 - Increased resistivity in P&G distribution network due to smaller feature sizes
 - Reduces the current carrying capabilities of P&G networks
 - High current transients in P&G buses
 - Voltage drops across the P&G networks
 - Can lead to incorrect logic operation
 - and/or reduction in switching speeds
 - Lifetime reductions and complete failure due to electromigration
 - Also contributes to an increase in dynamic power dissipation

PROBLEM SOLUTIONS

- **Accurate estimates of voltage drops and current densities in P&G buses**
 - Use circuit level analog simulators (like SPICE) for current and voltage estimations
 - The main problems are the Time and Computational cost due to size of P&G networks
- **Faster simulation times achieved by using**
 - custom tools targeted towards solving this specific problem
 - simplified device models
 - make use of tree or mesh structures of P&G networks
 - find reduced order approximation of the transfer functions
 - problems include loss of accuracy and algorithm stability
- **All solutions proposed so far are numeric in nature**

SOME EXISTING ALGORITHMS

- J. E. Hall, D. E. Hovevar, P. Yang, and M. J. McGraw, "SPIDER-- a CAD system for modeling VLSI metallization patterns," IEEE Transactions on CAD, Nov. 1987.
 - Estimates Median-Time-to-Failure for each section of a metal bus
 - Makes use of SPICE runs
 - User must provide estimates of current sources
 -
- S. Choudhury and M. A. Breuer, "Optimum design of IC power/ground nets subject to reliability constraints," IEEE Transactions on CAD, July 1988.
 - More general technique
 - Takes into consideration electromigration and voltage drops from leaf to node
 - Transform the problem into an unconstrained optimization problem

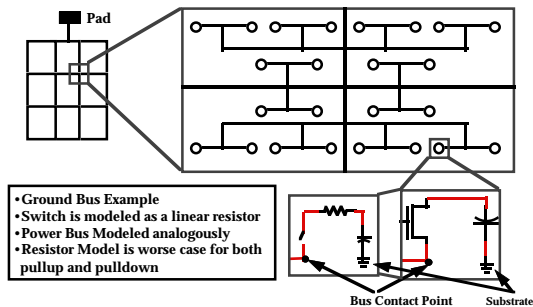
SOME EXISTING ALGORITHMS (cont ...)

- R. Dutta and M. Marek-Sadowska, "Automatic sizing of power/ground P/G networks in VLSI," Proceedings DAC, pp. 783-786, 1989.
 - Modeled as a nonlinear programming problem
 - Cost function is the sum of the segments areas
 - Constraints are wire width
 - Convergence is not necessary, meaningful results are achieved at any time the iterations are terminated
- T. Mitsuhashi and E. S. Kuh, "Power and ground network topology optimization for cell-based VLSIs," in Proceedings of the ACM/IEEE Design Automation Conference, pp. 524-529, 1992.

THIS ALGORITHM & SYMBOLIC ANALYSIS

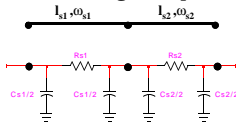
- **Motivation:** Use of SPICE is computationally expensive and other current methods use simplified current estimation techniques but are limited to performing DC analysis
- Sapatnekar and Shah presented a method (1996) that uses frequency domain techniques:
 - not limited to a resistive or any one class of networks
 - can perform transient analysis (with simplified models)
 - near linear algorithm (in the size of the network)
 - results within 1% of SPICE
- The algorithm is very suitable for symbolic analysis because of its linear models, frequency domain techniques and repetitive evaluations

P&G BUS MODEL



SEGMENTED RC MODEL of P&G BUS

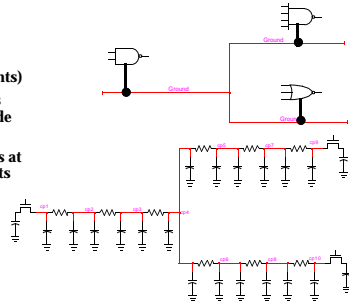
- Modeled as a set of segments connected in the form of a tree
- Each segment is modeled using a lumped RC
- π model
- $R_i = \rho l_i / \omega_i$
- $C_i = \beta l_i \omega_i$



- where l_i is the length of the i th segment, ω_i is the width of the i th segment, ρ is the sheet resistance and β is the wire capacitance per unit area

EXAMPLE MODEL FOR GROUND NET

- The P&G network is modeled as a tree T having $m+1$ nodes
- m edges (wire segments)
- one or more switches connected to each node
- Assume switches undergo state changes at predefined time points



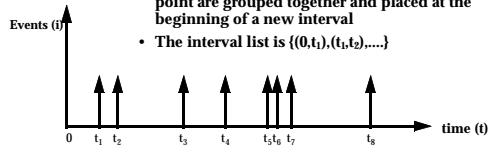
SYMBOLIC ANALYSIS APPLICATION

- The algorithm computes the time domain voltage waveform at each node for every time instance
- Two steps, which are repeated for every time point at which one or more switches changed states.
 - 1) Solve a set of equations of the form $(G + sC) V(s) = J(s)$ using an efficient path tracing algorithm, to obtain $V(s)$.
 - 2) Given $V(s)$ at every node, compute the time domain response by approximating $V(s)$ with a rational Padé approximation.

Both steps are perfectly suitable for symbolic analysis:
 Linear, fixed topology network that requires many computational iterations

ALGORITHM BASIS

- Event-driven algorithm where the events specified are the time points at which one or more switches change their state.
- The events are typically generated by an event-driven simulator
- All the switches that switch at the same time point are grouped together and placed at the beginning of a new interval
- The interval list is $\{(0,t_1),(t_1,t_2),\dots\}$



ALGORITHM OUTLINE

- Construct interval
- Select every element in the interval list in that order
- Set the corresponding switch states
- Process the interval (compute new states)
- Repeat with the the new switch states and computed initial conditions for next interval.
- Continue until all the intervals are processed.

COMPUTATION WITHIN AN INTERVAL

- The initial conditions on the capacitors are taken into consideration
- Compute the moments of each node voltage in the P&G networks
- Generate a stable Padé approximation
- Transform to a time domain voltage waveform
- The response is used to compute the initial conditions at all the nodes at the time of the next switching event
- Repeat the process

COMPUTATION WITHIN AN INTERVAL (continued...)

- S = Set of switches connected to the P&G bus
- $E = \{e_1, \dots, e_n\}$ is the set of all the segments in T
- $I = \{I_k \mid k = 1, \dots, I_{max}\}$ is the interval list
- I_k is the k^{th} interval
- Associated with each interval I is a subset of switches s_i and their states p_k
- t_k is the time span of interval I_k

```

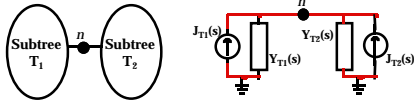
BEGIN ALGORITHM Vdrop()
I = IntervalList();
foreach  $I_k$  in I
  foreach  $j$  in  $s_k$  and  $state(j)$  in  $p_k$ 
    SetSwitch( $T, j, state(j)$ )
    PropagateActivity( $T, j, state(j)$ );
  foreach  $n$  in Nodes( $T$ )
     $f_n(t) = \text{TimeDomainResponse}(\text{moments}(n))$ ;
    SetInitialCondition( $n, f_n(t_k)$ )
END ALGORITHM Vdrop()
  
```

MOMENT COMPUTATION BY PATH TRACING

- Computation of moments of the voltage at any given node
- Recursively reduce the subtrees rooted at the given node n to an admittance $Y_{T1}(n)$ and a current source $J_{T1}(n)$

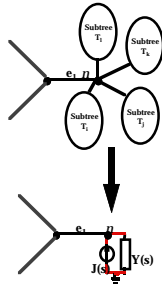
where

- $Y_{T1}(n)$ is the equivalent admittance of the subtree T_i , as seen from node n to the ground
- $J_{T1}(n)$ is the combined effect of all the switch elements in T_i , which have contributed to the current in the ground net.



MOMENT COMPUTATION (continued...)

- The equivalent current sources and admittances are computed for every node in linear time using a path tracing algorithm.
- The algorithm proceeds from the leaf-nodes, which are the contact points at the terminating segments of the ground net.
- A leaf-node could have multiple switches but exactly one segment connected to it.



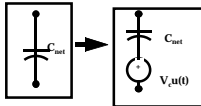
SYMBOLIC MODEL COMPUTATION

- $J(n)$ and $Y(n)$ can be computed symbolically
- It is the same as a symbolic implementation of Norton's theorem
- SCAPP used to generate a sequence of expressions
- A simpler symbolic implementation possible
- The equivalent current source of a single switch i is given by:

$$J^i = \frac{V_c}{s} \frac{sC_{net}}{1 + R_d s C_{net}}$$

SYMBOLIC MODEL COMPUTATION

- V_c is the initial voltage on the capacitor C_{net} at the start of the given interval.
- The pole at the origin in indicates that C_{net} is modeled as



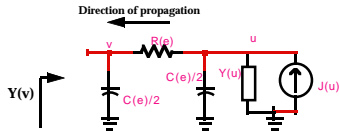
$$J^i = \frac{V_c}{s} \frac{sC_{net}}{1 + R_d s C_{net}}$$

SYMBOLIC MODEL COMPUTATION (continued...)

- Using Maclaurin series polynomial approximations we have $J^i(n) = V_c(sC_{net} - R_d s^2 C_{net}^2 + R_d s^3 C_{net}^3 + \dots)$
- Total admittance and equivalent current is: $Y^i = \frac{sC_{net}}{1 + R_d s C_{net}}$
- The expressions includes the effect of all the switches with binary operators controlling the closing and opening of the switches during the evaluation phase
 - $Y^i = sC_{net} - R_d s^2 C_{net}^2 + R_d s^3 C_{net}^3 + \dots$
 - $Y(n) = \sum J^i \forall i \in \text{Switches at } n$
 - $J(n) = \sum J^i \forall i \in \text{Switches at } n$

EFFECT OF THE CURRENT SOURCES

- The effect of the equivalent current sources is propagated to every node in the network
- Terminate segment e in the ground network, connecting a single contact point u to the rest of the network, rooted at v



EQUIVALENT ADMITTANCE

- The admittance and the equivalent current source, as seen into edge e from node v is computed as

$$Y(v) = \frac{1}{R(e) + (Y_0(u) + (Y_1(u) + 0.5C(e)) + \dots + Y_{2n-1}(u)s^{2n-1}) + 0.5C(e)}$$

$$= \frac{Y_0(u) + (Y_1(u) + 0.5C(e)) + \dots + Y_{2n-1}(u)s^{2n-1}}{1 + R(e)Y_0(u) + R(e)Y_1(u) + 0.5C(e) + \dots + (R(e)Y_{2n-1}(u)s^{2n-1}) + 0.5C(e)}$$

which yields

$$Y(v) = Y_0(v) + Y_1(v)s + Y_2(v)s^2 + \dots + Y_k(v)s^k + \dots$$

where

$$Y_k(v) = \frac{Y_k(u) - \sum_{i=1}^k R(e)Y_{k-i}(v)Y_i(u)}{1 + R(e)Y_0(u)}$$

EQUIVALENT CURRENT SOURCE

Using Norton's theorem (shorting node v to ground) we get

$$J_k(v) = J_0(v) + J_1(v)s + J_2(v)s^2 + \dots + J_k(v)s^k + \dots$$

where

$$J_k(v) = \frac{J_k(u) - \sum_{i=1}^k R(e)J_{k-i}(v)Y_i(u)}{1 + R(e)Y_0(u)}$$

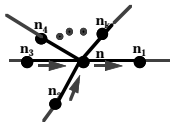
Thus, we say that the equivalent admittance and current source have been propagated from node u to v

GENERALIZING THE ALGORITHM

- The above case only considers the switches connected to leaf-nodes
- In general, switches can be connected to any node
- In cases where a node has more than one segment and multiple switches connected, the switches are processed in a similar manner
- The admittance and current source can be propagated if and only if a maximum of one segment is left unprocessed at the given node.

GENERAL NODE CASE

- Since each node n in a P&G network is a root of a subtree T_i , the equivalent devices seen through edge e_i can be computed by



$$J^{e_i}(n) = \sum_{j=1, j \neq i}^k J_{T_j}(n)$$

$$Y^{e_i}(n) = \sum_{j=1, j \neq i}^k Y_{T_j}(n)$$

GENERAL NODE CASE (continued...)

- Voltage moments at n :

$$V(n) = \frac{J(n)}{Y(n)} \quad \text{where} \quad \begin{aligned} J(n) &= \sum_{i=1}^k J_{T_i}(n) \\ Y(n) &= \sum_{i=1}^k Y_{T_i}(n) \end{aligned}$$

Moments of $V(n)$:

$$V(n) = V_0(n) + V_1(n)s + V_2(n)s^2 + \dots + V_k(n)s^k + \dots$$

$$\text{where } V_k(n) = \frac{J_k(n) - \sum_{i=1}^k V_{k-i}(n)Y_i(n)}{Y_0(n)}$$

GENERAL NODE CASE (continued...)

- The voltage moments can be computed at all nodes if equivalent Ys and Js have been propagated twice along each segment: once in each direction. Hence a linear time path tracing algorithm which traces each segment once in either direction can be used for the computation of voltage moments at all nodes.

COMPUTING TIME DOMAIN RESPONSE

- Use moment matching techniques
 - We have V(s) for each node in the network
 - 2N moments for the node voltages have been computed
 - Largest order of approximation can be N, let M=N-1
 - Approximate V(s) with an M-zero N-pole expression
 - Padé Approximation

$$\tilde{V}(s) = \frac{a'_0 + a'_1 s + a'_2 s^2 + \dots + a'_M s^M}{1 + b'_1 s + b'_2 s^2 + \dots + b'_N s^N}$$

TIME DOMAIN (continued ..)

- The denominator coefficients are computed from the Hankel matrix

$$\begin{bmatrix} V_0 & V_1 & \dots & V_N \\ V_1 & V_2 & \dots & V_{N+1} \\ \vdots & \vdots & \dots & \vdots \\ V_N & V_{N+1} & \dots & V_{N+M-1} \end{bmatrix} \begin{bmatrix} b'_N \\ b'_{N-1} \\ \vdots \\ b'_1 \end{bmatrix} = - \begin{bmatrix} V_{N+1} \\ V_{N+2} \\ \vdots \\ V_{N+M} \end{bmatrix}$$

The numerator coefficients are computed by equating powers of s

$$\begin{aligned} a'_0 &= V_0 \\ a'_1 &= V_1 + b'_1 V_0 \\ &\vdots \\ a'_M &= V_M + \sum_{i=1}^M b'_i V_{M-i} \end{aligned}$$

TIME DOMAIN (cont ...)

- Given the above symbolic coefficients, the time domain response can be found numerically by writing

$$\tilde{V}(s) = \hat{c} + \sum_{i=1}^N \frac{\hat{k}_i}{s - \hat{p}_i}$$

- which is transformed to the time domain as

$$\tilde{v}(t) = \hat{c}\delta(t) + \sum_{i=1}^N \hat{k}_i e^{\hat{p}_i t}$$

PRELIMINARY EXPERIMENTAL RESULTS

- Two randomly generated ground nets
 - gnet1: 400 contact points
total wire length=2cm
wire width=3μ
 - gnet2: 4000 contact points
total wire length=10cm
wire width=3μ
- Accuracy was compared with HSPICE at four selected nodes A, B, C and D
- Comparing peak voltage and root mean square error
- Sample time points were 0.01ns
- Maximum order of the Padé Approximations was set to 3

PRELIMINARY EXPERIMENTAL RESULTS (continued ...)

- gnet1: Peak and RMSE Voltages**

Node	Peak (Volts) Vdrop	Peak (Volts) HSPICE	VRMSE (Volts)
A	0.350	0.342	0.00130
B	0.432	0.426	0.00150
C	0.688	0.689	0.00087
D	0.461	0.463	0.00054

- gnet2: Peak and RMSE Voltages**

Node	Peak (Volts) Vdrop	Peak (Volts) HSPICE	VRMSE (Volts)
A	0.216	0.229	0.00017
B	0.225	0.243	0.00013
C	0.419	0.419	0.00028
D	0.129	0.125	0.00039

CONCLUSIONS

- Expand the applications of symbolic analysis to a new field
- Application of symbolic analysis to Power and Ground interconnect analysis
- Provides mechanism for reducing cost of highly iterative process
- More testing of the algorithm in general and the symbolic role in particular is required
- Expand algorithm to perform P&G bus optimization, ie. produce an optimal P&G network based on the results