

Nonlinear Systems

Lecture 19

04/04/13

Last time:

→ Example of MRAC

$$\dot{y} = ay + u \quad (1)$$

a : constant but unknown

MRAC: Adaptive controller that makes (1) behave like a reference model

$$\dot{y}_m = -a_m y_m + r$$

$\left\{ \begin{array}{l} \downarrow \\ a_m > 0 \end{array} \right.$ $\left. \begin{array}{l} \downarrow \\ \text{reference signal} \end{array} \right.$

→ Integrator backstepping

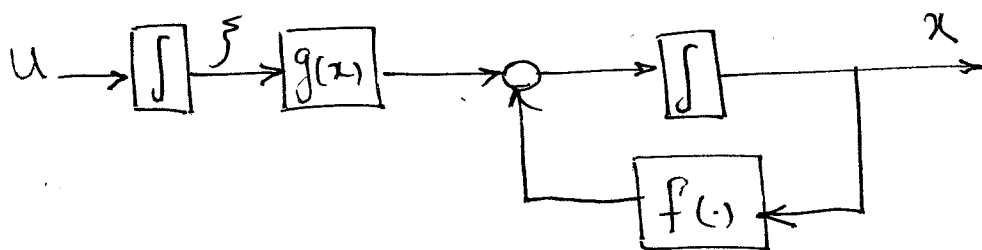
$$\dot{x} = f(x) + g(x)\xi \quad (1) \text{ (affine dependence on } \xi \text{)}$$

$$ax + b \quad \text{vs} \quad ax$$

(affine) (linear)

$$\dot{\xi} = u$$

(2)



If ξ was control ($\xi = \alpha(x)$) and guarantees global asymptotic stability (GAS) of $\bar{x} = 0$ for (1) with a Lyapunov function $V(x)$

$$\frac{\partial V}{\partial x} [f(x) + g(x)\alpha(x)] < 0, \quad \forall x \neq 0$$

then backstepping tells you how to design

$u = \alpha_2(x, \xi)$ that provides GAS of

$(\bar{x}, \bar{\xi}) = (0, 0)$ of (1)-(2).

Ex

$$\dot{x}_1 = x_1^2 + x_2$$

$$\dot{x}_2 = u$$

step 1: consider $\dot{x}_1 = x_1^2 + x_2$ and think of x_2 as a control.

$$\text{Propose } V_1(x_1) = \frac{1}{2} x_1^2$$

$$\dot{V}_1 = x_1 \dot{x}_1 = x_1 (x_1^2 + x_2)$$

Eg. $x_2 = -x_1^2 - K_1 x_1$ $K_1 > 0$

or

$$x_2 = -x_1^2 - K_1 x_1^3 \quad K_1 > 0$$

or

⋮

Abstractly : $x_2 = \alpha(x_1)$

If x_2 was control \Rightarrow done!

But it is not \Rightarrow need to account for the fact that x_2 is ~~not~~ a state variable (not control)

step 2: $z_2 = x_2 - \alpha(x_1)$

$$\dot{z}_2 = \dot{x}_2 - \dot{\alpha}(x_1)$$

$$= u - \frac{\partial \alpha}{\partial x_1} \dot{x}_1 = u - \frac{\partial \alpha}{\partial x_1} [x_1^2 + x_2]$$

$$= u - \frac{\partial \alpha}{\partial x_1} [x_1^2 + z_2 + \alpha(x_1)]$$

so we introduced the new variable z_2 , found an equation for it.

\Rightarrow Augment Lyapunov function from step (1):

$$V_a(x_1, z_2) = V_1(x_1) + \frac{1}{2} z_2^2 = \frac{1}{2} x_1^2 + \frac{1}{2} z_2^2$$

$$\begin{aligned}\dot{V}_a &= \dot{V}_1 + z_2 \dot{z}_2 = x_1(x_1^2 + \alpha(x_1) + z_2) + z_2(u - \dot{\alpha}) \\ &= x_1(x_1^2 + \alpha(x_1)) + z_2(u - \dot{\alpha} + x_1) \\ &\quad \underbrace{\hspace{10em}}_{-W_1(x_1)}\end{aligned}$$

Choose:

$$\boxed{u = -x_1 + \dot{\alpha} - K_2 z_2} \quad (*)$$

\uparrow
 $K_2 > 0$

to obtain: $\dot{V}_a = -W_1(x_1) - K_2 z_2^2 < 0$

GAS of the origin of (1)-(2).

Note! No reason to explicitly differentiate α in the expression for control (*). Use instead the analytical expression for $\dot{\alpha}$:

$$\frac{\partial \alpha}{\partial t} = \frac{\partial \alpha}{\partial x_1} \dot{x}_1 = \frac{\partial \alpha}{\partial x_1} [x_1^2 + x_2]$$

Thus u given by (*) is a static nonlinear state feedback control law, which provides GAS. ($u = \beta(x_1, x_2)$)

Therefore backstepping provided a family of control laws in a constructive manner that achieve GAS.

Note! Backstepping yields:

→ A family of globally stabilizing control laws. These families are parameterized by different choices of stabilizing functions.

→ Control Lyapunov Function (CLF), which can be used to obtain additional control laws. (later) ✓

Additional comments:

$$\begin{cases} \dot{x} = f(x) + g(x) \xi \\ \dot{\xi} = f_a(x, \xi) + g_a(x, \xi) u \end{cases}$$

! $g_a(x, \xi) \neq 0$ for all x, ξ (same as before)

$$u = \frac{1}{g_a(x, \xi)} (\dots)$$

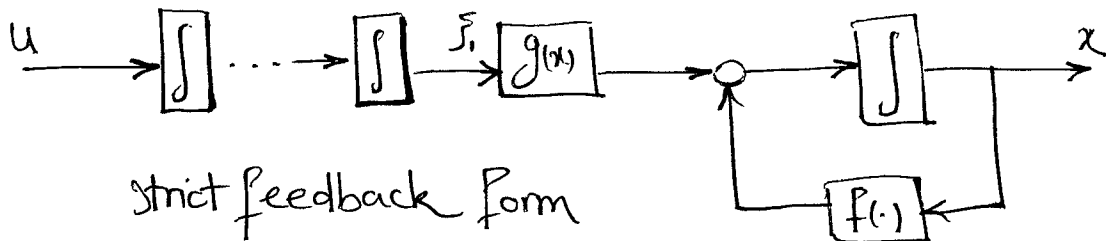
Ex $\dot{x}_1 = x_1^2 + x_2$

$$\dot{x}_2 = x_3$$

$$\dot{x}_3 = u$$

Apply Backstepping.

$$z_3 = x_3 - \alpha_2(x_1, x_2) \quad \text{and ... (same as before)}$$



$$\dot{x}_1 = f_1(x_1) + g_1(x_1) \underline{x_2}$$

$$\dot{x}_2 = f_2(x_1, x_2) + g_2(x_1, x_2) \underline{x_3}$$

$$\dot{x}_3 = f_3(x_1, x_2, x_3) + g_3(x_1, x_2, x_3) \underline{u}$$

⋮

Ex $\dot{x}_1 = x_1^2 + x_2$

$$\dot{x}_2 = u$$

Objective: $x_1(t) \xrightarrow{t \rightarrow \infty} r(t)$

$$y_r(t)$$

$$z_1 := x_1 - r$$

$$y_d(t)$$

$$\begin{aligned}\dot{z}_1 &= \dot{x}_1 - \dot{r} = x_1^2 + x_2 - \dot{r} \\ &= (z_1 + r)^2 + x_2 - \dot{r}\end{aligned}$$

⋮

$$V_1(z_1) = \frac{1}{2} z_1^2$$

(Needs to know reference signal and its derivative)

Control Lyapunov Function (CLF)

$$\dot{x} = f(x) + g(x) \cdot u$$

$u = \alpha(x)$: control law that gives ↓

$$\underbrace{\frac{\partial V}{\partial x} f(x)}_{L_f V(x)} + \underbrace{\frac{\partial V}{\partial x} g(x) \alpha(x)}_{L_g V(x)} < 0$$

$$L_f V(x) + L_g V(x) \alpha(x) < 0$$

If there are states x st. $L_g V(x) = 0$ and $L_f V(x) < 0$ at these points then control does not enter into equations for the derivative of V .

Propose $V(x)$:

$$\dot{V}|_{(1)} = \frac{\partial V}{\partial x} [f(x) + g(x)u] = L_f V(x) + \underbrace{L_g V(x) \cdot u}_{\text{when } = 0} \Rightarrow L_f V(x) < 0$$

Def: A smooth positive definite radially unbounded function $V(x)$ is called a CLF for the system $\dot{x} = f(x) + g(x)u$ if for all $x \neq 0$ $L_g V(x) = 0 \Rightarrow L_f V(x) < 0$.

There are many ways to generate control laws from CLF:

E.g. Sontag's Formula (Eduardo Sontag, Rutgers, NJ)

$$u(x) = \begin{cases} 0 & , \quad L_g V(x) = 0 \\ -\frac{L_f V(x) + \sqrt{(L_f V(x))^2 + (L_g V(x))^2}}{L_g V(x)} & \text{o.w.} \end{cases}$$

Comment!

This control law is continuous if the CLF satisfies the "small control property"

$\forall \epsilon > 0, \exists \delta > 0$ st. for each $x \neq 0$ with $\|x\| \leq \delta$

\Rightarrow we can find u with $\|u\| \leq \epsilon$ st.

$$L_f V(x) + L_g V(x) u < 0$$