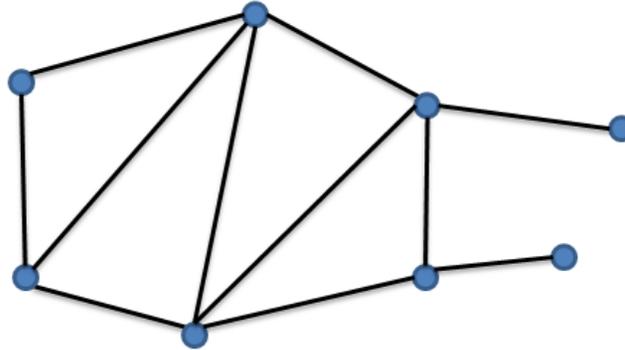


Lecture 27: Optimal control of undirected graphs



- Single-integrator dynamics

$$\dot{x}_i = u_i + d_i$$

- Relative information exchange with neighbors

$$u_i(t) = - \sum_{j \in \mathcal{N}_i} k_{ij} (x_i(t) - x_j(t))$$

- Closed-loop dynamics

$$\dot{x}(t) = - L(k) x(t) + d(t)$$

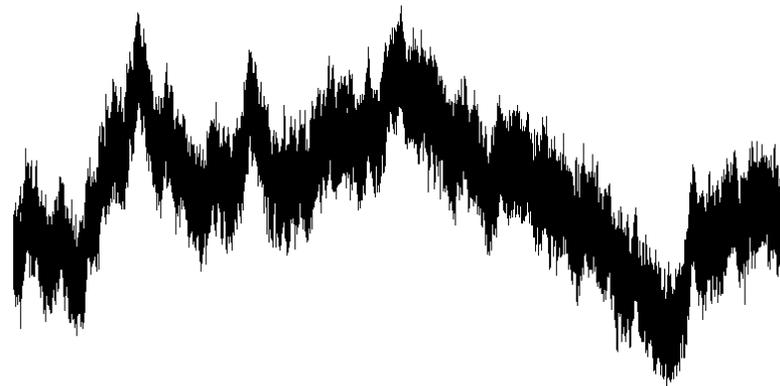
- Structured matrix L depends on $\left\{ \begin{array}{l} \text{graph topology} \\ \text{vector of feedback gains } k \end{array} \right.$

- Independent of graph topology and feedback gains

$$L(k) \mathbf{1} = 0 \cdot \mathbf{1}$$

Average mode

$$\bar{x}(t) = \frac{1}{N} \sum_{i=1}^N x_i(t) : \text{undergoes random walk}$$



If other modes are stable, $x_i(t)$ fluctuates around $\bar{x}(t)$

deviation from average: $\tilde{x}_i(t) = x_i(t) - \bar{x}(t)$

steady-state variance: $\lim_{t \rightarrow \infty} \mathcal{E} (\tilde{x}^T(t) \tilde{x}(t))$

Optimal control problem

What graph topologies lead to small variance?

How to design feedback gains to minimize variance?

$$\dot{x}(t) = -L(k)x(t) + d(t)$$

$$z(t) = \begin{bmatrix} \tilde{x}(t) \\ u(t) \end{bmatrix} = \begin{bmatrix} I - \frac{1}{N} \mathbf{1}\mathbf{1}^T \\ -L(k) \end{bmatrix} x(t)$$

- Setup:

- ★ Undirected graphs: bi-directional interaction between nodes
- ★ Symmetric feedback gains

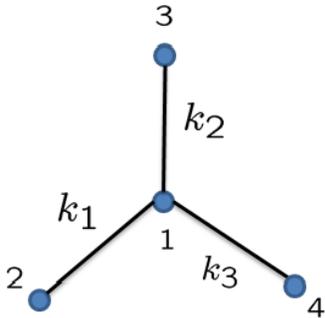
$$k_{ij} = k_{ji} \Rightarrow L(k) = L^T(k)$$

Incidence matrix

- Edge $l \sim (i, j)$: connects nodes i and j
- ★ Define $e_l \in \mathbb{R}^N$ with only two nonzero entries

$$(e_l)_i = 1 \quad (e_l)_j = -1$$

$$\text{Incidence matrix: } E = [e_1 \cdots e_m]$$



$$E = \begin{bmatrix} 1 & 1 & 1 \\ -1 & 0 & 0 \\ 0 & -1 & 0 \\ 0 & 0 & -1 \end{bmatrix}, \quad E^T x = \begin{bmatrix} x_1 - x_2 \\ x_1 - x_3 \\ x_1 - x_4 \end{bmatrix}, \quad E^T \mathbf{1} = 0$$

Edge $l \sim (i, j)$: $k_l := k_{ij} = k_{ji}$

$$\text{Laplacian: } L(K) = E K E^T = \sum_{l=1}^m k_l e_l e_l^T$$

$$\text{Structured feedback gain: } K = \begin{bmatrix} k_1 & & \\ & \ddots & \\ & & k_m \end{bmatrix}$$

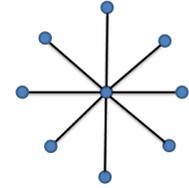
Tree graphs

- Trees: connected graphs with no cycles

path



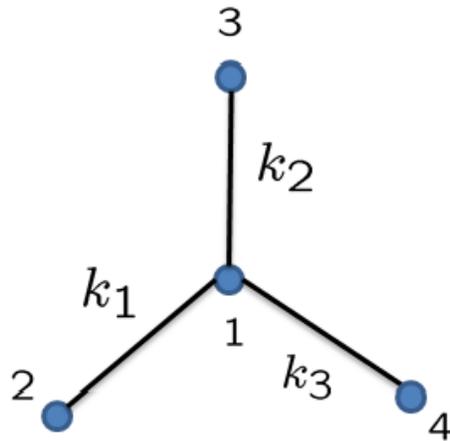
star



Incidence matrix of a tree $E_t \in \mathbb{R}^{N \times (N-1)}$



$$E_t = \begin{bmatrix} 1 & 0 & 0 \\ -1 & 1 & 0 \\ 0 & -1 & 1 \\ 0 & 0 & -1 \end{bmatrix}$$



$$E_t = \begin{bmatrix} 1 & 1 & 1 \\ -1 & 0 & 0 \\ 0 & -1 & 0 \\ 0 & 0 & -1 \end{bmatrix}$$

- Coordinate transformation

$$\begin{bmatrix} \psi(t) \\ \bar{x}(t) \end{bmatrix} = \underbrace{\begin{bmatrix} E_t^T \\ \frac{1}{N} \mathbf{1}^T \end{bmatrix}}_T x(t) \Leftrightarrow x(t) = \underbrace{\begin{bmatrix} E_t (E_t^T E_t)^{-1} & \mathbf{1} \end{bmatrix}}_{T^{-1}} \begin{bmatrix} \psi(t) \\ \bar{x}(t) \end{bmatrix}$$

In new coordinates

$$\begin{aligned} \begin{bmatrix} \dot{\psi}(t) \\ \dot{\bar{x}}(t) \end{bmatrix} &= - \begin{bmatrix} E_t^T \\ \frac{1}{N} \mathbf{1}^T \end{bmatrix} E_t K E_t^T \begin{bmatrix} E_t (E_t^T E_t)^{-1} & \mathbf{1} \end{bmatrix} \begin{bmatrix} \psi(t) \\ \bar{x}(t) \end{bmatrix} + \begin{bmatrix} E_t^T \\ \frac{1}{N} \mathbf{1}^T \end{bmatrix} d(t) \\ &= \begin{bmatrix} -E_t^T E_t K & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} \psi(t) \\ \bar{x}(t) \end{bmatrix} + \begin{bmatrix} E_t^T \\ \frac{1}{N} \mathbf{1}^T \end{bmatrix} d(t) \\ z(t) &= \begin{bmatrix} I - \frac{1}{N} \mathbf{1} \mathbf{1}^T \\ -E_t K E_t^T \end{bmatrix} \begin{bmatrix} E_t (E_t^T E_t)^{-1} & \mathbf{1} \end{bmatrix} \begin{bmatrix} \psi(t) \\ \bar{x}(t) \end{bmatrix} \end{aligned}$$

Tree graphs: structured optimal H_2 design

$$\dot{\psi}(t) = -E_t^T E_t K \psi(t) + E_t^T d(t)$$

$$z(t) = \begin{bmatrix} E_t (E_t^T E_t)^{-1} \\ -E_t K \end{bmatrix} \psi(t)$$

H_2 norm (from d to z)

$$J(K) = \frac{1}{2} \text{trace} \left((E_t^T E_t)^{-1} K^{-1} + K E_t^T E_t \right)$$

Diagonal matrix: $K = \begin{bmatrix} k_1 & & \\ & \ddots & \\ & & k_{N-1} \end{bmatrix}$

- Structured optimal feedback gains

$$k_i = \sqrt{\frac{(E_t^T E_t)^{-1}_{ii}}{2}}, \quad i = 1, \dots, N-1$$

- In Lecture 28, I made a blunder on board while deriving the optimal values of k_i

Here is correct derivation:

★ $G := (E_t^T E_t)^{-1} \Rightarrow$ diagonal elements of G determined by $G_{ii} = (E_t^T E_t)^{-1}_{ii}$

★ All diagonal elements of $E_t^T E_t$ are equal to 2

$$\begin{aligned} E_t^T E_t &= [e_1 \cdots e_{N-1}]^T [e_1 \cdots e_{N-1}] = \begin{bmatrix} e_1^T \\ \vdots \\ e_{N-1}^T \end{bmatrix} [e_1 \cdots e_{N-1}] \\ &= \begin{bmatrix} e_1^T e_1 & \cdots & e_1^T e_{N-1} \\ \vdots & \ddots & \vdots \\ e_{N-1}^T e_1 & \cdots & e_{N-1}^T e_{N-1} \end{bmatrix} = \begin{bmatrix} 2 & \cdots & e_1^T e_{N-1} \\ \vdots & \ddots & \vdots \\ e_{N-1}^T e_1 & \cdots & 2 \end{bmatrix} \end{aligned}$$

★ K – diagonal matrix $\Rightarrow J(K)$ can be written as

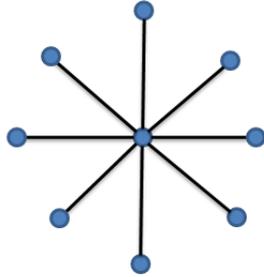
$$J(K) = \sum_{i=1}^{N-1} \left(\frac{G_{ii}}{2k_i} + k_i \right)$$

★ $J(K)$ in a separable form \Rightarrow element-wise minimization will do

$$\frac{d}{dk_i} \left(\frac{G_{ii}}{2k_i} + k_i \right) = -\frac{G_{ii}}{2k_i^2} + 1 = 0 \Rightarrow k_i = \sqrt{\frac{G_{ii}}{2}}, \quad i = 1, \dots, N-1$$

Optimal gains for star and path

• Star:

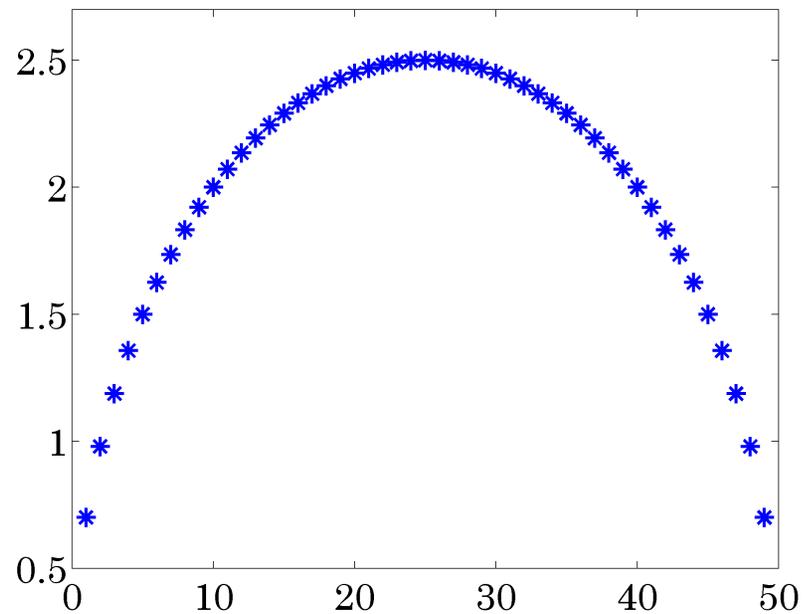


uniform gain $k = \sqrt{\frac{N-1}{2N}} \approx \frac{1}{\sqrt{2}}$ for large N

• Path:



$$k_i = \sqrt{\frac{i(N-i)}{2N}}$$



largest gains in the center

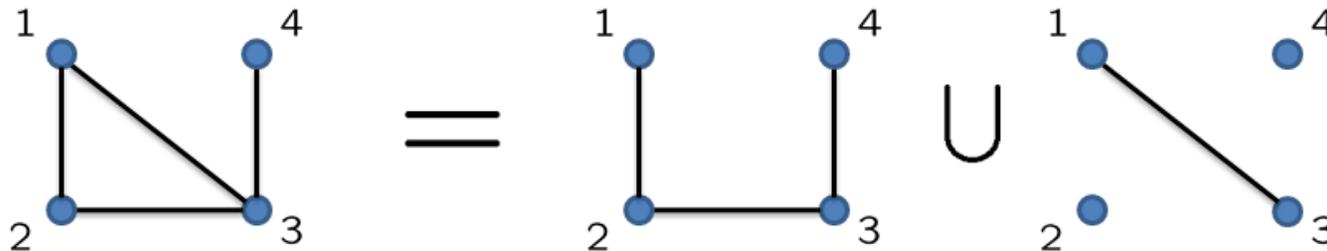
General undirected graphs

- Decompose graph into a tree subgraph and remaining edges

Incidence matrix: $E = [E_t \ E_c]$

Projection matrix: $\Pi = E_t E_t^+ = E_t (E_t^T E_t)^{-1} E_t^T$

$E_c \in \text{range}(\Pi)$: $E_c = \Pi E_c$



$$\begin{bmatrix} 1 & 0 & 0 & 1 \\ -1 & 1 & 0 & 0 \\ 0 & -1 & 1 & -1 \\ 0 & 0 & -1 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ -1 & 1 & 0 \\ 0 & -1 & 1 \\ 0 & 0 & -1 \end{bmatrix} \cup \begin{bmatrix} 1 \\ 0 \\ -1 \\ 0 \end{bmatrix}$$

$$\begin{aligned} E &= [E_t \ E_c] = [E_t \ \Pi E_c] \\ &= E_t [I \ (E_t^T E_t)^{-1} E_t^T E_c] = E_t M \end{aligned}$$

General graphs: structured optimal H_2 design

$$\dot{\psi}(t) = -E_t^T E_t M K M^T \psi(t) + E_t^T d(t)$$

$$z(t) = \begin{bmatrix} E_t (E_t^T E_t)^{-1} \\ -E_t M K M^T \end{bmatrix} \psi(t)$$

tree graphs: $M = I$

H_2 norm (from d to z)

$$J(K) = \frac{1}{2} \text{trace} \left((E_t^T E_t)^{-1} (M K M^T)^{-1} + M K M^T E_t^T E_t \right)$$

- Main result:

- ★ Closed-loop stability $\Leftrightarrow M K M^T > 0$

$\{W_1 > 0, W_2 = W_2^T\}$ then $-W_1 W_2$ Hurwitz $\Leftrightarrow W_2 > 0$

- ★ $M K M^T > 0 \Rightarrow$ convexity of $J(K)$

- Semi-definite program

$$\text{minimize } \frac{1}{2} \text{trace} (X + M K M^T E_t^T E_t)$$

$$\text{subject to } \begin{bmatrix} X & (E_t^T E_t)^{-1/2} \\ (E_t^T E_t)^{-1/2} & M K M^T \end{bmatrix} > 0$$

K diagonal

- Use CVX to solve it

```

cvx_begin sdp

    variable k(Ne) % vector of unknown feedback gains

    variable X(Nv-1,Nv-1) symmetric;
    X == semidefinite(Nv-1); % Schur complement variable

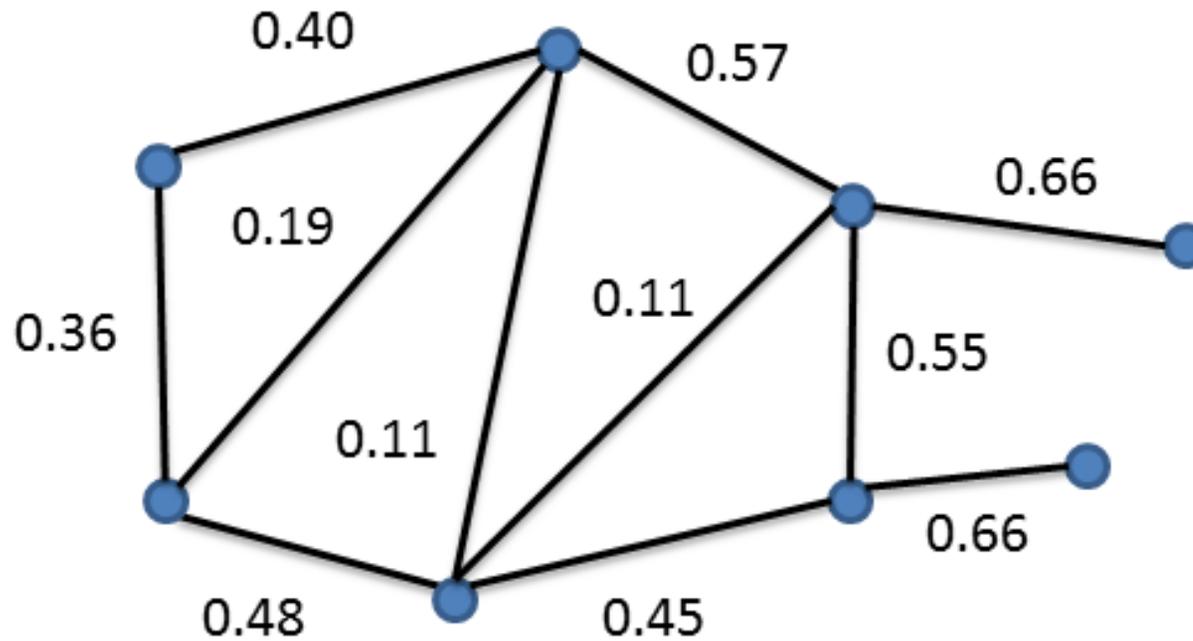
    Mk = M*diag(k)*M'; % Matrix Mk

    minimize(0.5*trace( q*X + r*Mk*W ))
    subject to [X, invWh; invWh, Mk] > 0;

cvx_end

```

Examples

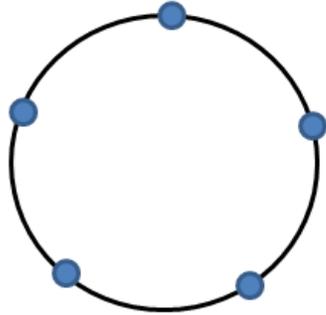


- Compare with performance of uniform gain design

J^*	$J(k = 1)$	$(J - J^*)/J^*$
9.1050	13.1929	45%

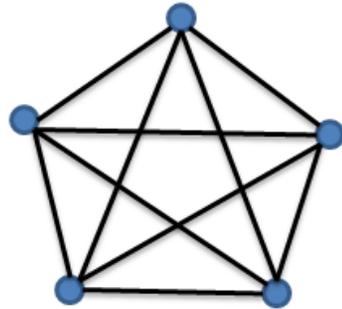
- Analytical results for circle and complete graphs

★ Circle



uniform gain $k = \sqrt{\frac{N^2 - 1}{24N}}$

★ Complete graph



uniform gain $k = \frac{2}{N}$

Additional material

- Papers to read
 - ★ *Xiao, Boyd, Kim, J. Parallel Distrib. Comput. '07*
 - ★ *Zelazo & Mesbahi, IEEE TAC '11*
 - ★ *Lin, Fardad, Jovanovic, CDC '10*